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## **OPEN** Broadband and Broad-angle **Polarization-independent Metasurface for Radar Cross Section Reduction**

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In this work, a broadband and broad-angle polarization-independent random coding metasurface structure is proposed for radar cross section (RCS) reduction. An efficient genetic algorithm is utilized to obtain the optimal layout of the unit cells of the metasurface to get a uniform backscattering under normal incidence. Excellent agreement between the simulation and experimental results show that the proposed metasurface structure can significantly reduce the radar cross section more than 10 dB from 17 GHz to 42 GHz when the angle of incident waves varies from 10° to 50°. The proposed coding metasurface provides an efficient scheme to reduce the scattering of the electromagnetic waves.

Engineered metamaterials and metasurfaces usually comprise periodic or non-periodic subwavelength unit cells that enable the manipulation of electromagnetic wave propagation to obtain unusual properties<sup>1-3</sup> such as negative refraction<sup>4-6</sup>, subwavelength focusing<sup>7-9</sup>, and electromagnetic invisibility cloaking<sup>10-16</sup>. Among these examples, electromagnetic invisibility cloaking represents a prominent utilization of metamaterials for greatly reducing the radar cross section (RCS) of targets in military applications<sup>17,18</sup>. A perfect invisibility cloak presents no backward or forward RCS, and the optical transformation method provides an efficient means of bending incident electromagnetic waves along a given region for this purpose<sup>19-21</sup>. A chessboard-like structure combining perfect electric conductors and artificial magnetic conductors (AMCs) has been employed to reduce the RCS effectively<sup>22</sup>. However, it is difficult to increase the bandwidth of AMC structures due to AMC resonance. Therefore, broadening the working bandwidth of a metasurface for RCS reduction is a major issue that should be addressed. Another method of reducing the RCS is scattering cancellation. In this case, the metamaterial is homogeneous, and leads to the polarization of the incident wave converted to its cross-polarized one, which represents the plasmon cloaking of an object<sup>23-25</sup>. Recently, a new approach to reducing the scattering of electromagnetic waves using coding metasurfaces composed of digital coding elements has been proposed, and a 1-bit, 2-bit, and 3-bit random coding metamaterial has been reported<sup>26</sup>. Far-field scattering pattern analysis in conjunction with a hybrid optimization algorithm has been employed to obtain an optimal arrangement of digital coding elements for designing a low RCS metasurface with ultra-low backward scattering that is functional over a broadband from 7 GHz to 14 GHz<sup>27</sup>. In addition, the particle-swarm optimization algorithm has been employed to establish the optimal coding sequences of Minkowski closed loop particles for reducing the scattering of terahertz waves<sup>28</sup>.

Genetic algorithms represent another well established optimization method employed in electromagnetic applications. A GA is an optimization technique that searches for optimal solutions by simulating natural selection processes and the genetic mechanism of the biological evolutionary process. The technique was first proposed by Holland in 1975<sup>29</sup>. As early as the 1990s, Haupt<sup>30,31</sup> significantly employed GAs in electromagnetic research. Twenty years later, Johnson<sup>32</sup> employed a GA to optimize the design of one-dimensional (1D) and two-dimensional (2D) array antennas, and the optimization process was shown to overcome the limitations associated with conventional optimization techniques when applied to antenna arrays. Marcano<sup>33-37</sup> applied GAs in the design of linear and plane array antennas. Application of GAs for implementing RCS reduction has also been pursued. For example, Mosallaei<sup>38</sup> combined a GA with the mode equation, obtained in a wide frequency band,

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**Figure 1.** Front views of "0" and "1" element unit cells. The unit cell structure with an angle of 45° to the y axis represents the "0" element (a), and the structure rotated 90° counter-clockwise about the z axis is the "1" element (b). Reflection properties of "0" and "1" elements versus frequency under normal x-polarized electromagnetic incident waves: (c) lattice reflection amplitudes; (d) lattice reflection phases.



Figure 2. Simulated reflections of the unit cell under normal x- and y-polarized electromagnetic incident waves.

and applied the method to effectively minimize the RCS of a radar absorbing material (RAM). Zhu<sup>39</sup> combined a standard GA with the high frequency method to optimize RCS reduction by changing the normal anisotropy surface impedance distribution of a geometric structure for missile applications. Yao<sup>40</sup> obtained the dimensions of a metallic patch by means of GA optimization, and successfully designed a low RCS patch antenna. Thus, the application of GAs to electromagnetic research is quite mature, and the use of a GA to design a low RCS antenna array is certainly feasible.

In the present study, a random 1-bit digital coding metasurface for RCS reduction is designed based on diffuse reflection theory, and a wideband property is imparted to the metasurface by means of wideband cross polarization conversion<sup>41</sup>. The proposed metasurface exhibits broadband, broad-angle, and polarization-independent for RCS reduction. Rather than employing multi-bit elements, the necessary phase difference can be obtained with 1-bit elements by simply rotating the structures, and the proposed metasurface therefore consists of unit cells with "0" and "1" elements owing to their **0** and  $\pi$  phase responses. Furthermore, an efficient strategy for designing random coding metasurfaces is developed by employing a combination of diffuse reflection theory and the scattering pattern reshaping method in conjunction with a genetic algorithm (GA) to optimize the sequences of the unit cells of the non-periodic random metasurface structure. Both simulation and experimental measurement results indicate that the optimized random coding metasurface reduces the RCS by greater than 10 dB over a frequency



Figure 3. Flowchart of the genetic algorithm employed to search for the optimal unit cell configuration.

range of 17–42 GHz for both *x*- and *y-polarized* normal and obliquely incident electromagnetic waves, and the bandwidth is not reduced at all under oblique incidence. Both experimental and simulation results verify the reliability and validity of the proposed method.

### Results

Unit cell properties and optimal sequencing of the metasurface. In general, the unit cell structure of a polarization conversion metasurface exhibits different phase responses under normal incidence (along the *z-direction*) depending on the rotation angle in the *xy-plane*. Two structures with relative phase responses of  $\pi$  can be arranged in a chessboard-like metasurface to reduce the RCS under monostatic backscattering conditions. In the worst case, the reflection consists of four strong lobes, which can be easily detected by bistatic detection. A 1-bit digital coding metasurface was proposed by Cui et al. in 2014<sup>26</sup>, which included two types of unit cells with  $\boldsymbol{0}$  and  $\boldsymbol{\pi}$  phase responses to mimic the "0" and "1" elements, respectively, and RCS reduction was achieved through the arbitrary arrangement of the unit cells. Inspired by this concept, we designed a new structure by simply rotating the original structure by 90°, which also exhibits broadband, broad-angle, and polarization-independent characteristics. The original structure and the new structure are illustrated in Fig. 1(a,b), respectively, in which the F4B-2 woven glass polytetrafluoroethylene dielectric substrate has a thickness d = 1.5 mm with a dielectric constant  $\varepsilon_r = 2.65$  and  $\tan \delta = 0.001$ . The geometrical dimensions of the structure are a = 3 mm,  $w = 0.2 \text{ mm}, l = 2.6 \text{ mm}, b = 1.5 \text{ mm}, c = 0.45 \text{ mm}, h = 0.6 \text{ mm}, \alpha = 75^\circ, \beta = 90^\circ$  and the thickness of the metallic coating (PEC) t = 0.018 mm. The polarization conversion mechanism owing to the multiple resonances of the unit cells, and the reflection phase difference between the "0" and "1" unit cells remains  $\pi$  over the entire working band. Figure 2 illustrates the simulated reflection of the unit cell for normal x- and y-polarized incidence, in which  $R_{xx}/R_{yy}$  and  $R_{yx}/R_{xy}$  represent the reflection coefficients of the co-polarized and cross-polarized waves, respectively. Therefore, according to the work of Cui et al.<sup>28</sup>, we can easily design the coding metasurface using only two elements. To satisfy the periodic boundary required for element simulation, a lattice of dimension D, containing  $6 \times 6$  unit cells of equivalent orientation, is generated as the "0" or "1" element for the metasurface.

The simplest method of constructing an RCS-reducing metasurface using 1-bit unit cells is to generate a phase-distribution matrix with "0" and "1" elements randomly distributed, and place each element according to its reflection phase. However, a randomly constructed matrix does not necessarily exhibit the optimal phase distribution, and the resulting metasurface is not necessarily the best possible configuration. Hence, we adopt a GA to obtain the optimal configuration from a random sequence, and to ensure that the final design has the desired RCS reduction characteristics. We consider an  $M \times N$  dimensional random permutation matrix containing digital 0 and 1 components, which represent the "0" and "1" lattices with the respective reflection phases **0** and  $\pi$ . According to array theory, the lattice-scattered electric-field intensity in the far-field region can be expressed as



**Figure 4.** The normalized field intensity patterns obtained at 25.6 GHz for coding metasurfaces with coding sequences (**a**) 000000.../000000... (or 111111.../111111..., which yields equivalent results), (**b**) 010101.../ 010101..., (**c**) 010101.../101010..., and (**d**) the coding sequence obtained by the proposed optimization procedure. We see that the main lobe in the case of (**d**) is obviously decreased, and has produced many side lobes.



**Figure 5.** The optimized 1-bit coding metasurface of planform obtained by the GA. The "0" and "1" lattices comprise  $6 \times 6$  equivalent unit cells, and are distinguished by yellow and blue colors, respectively.

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$$E_{m,n} = K_{m,n} \Gamma_{m,n} f_{m,n}(\theta, \varphi) \frac{\exp(-jk_0 r_{m,n})}{r_{m,n}},$$
(1)

where, for m = 1, ..., M and n = 1, ..., N,  $K_{m,n}$  is the scale coefficient,  $r_{m,n}$  is the distance between the lattice point (m, n) and the far-field region observation point, and  $f_{m,n}(\theta, \varphi)$  is the lattice scattering pattern function, given with respect to the elevation and azimuth angles  $\theta$  and  $\varphi$ . Moreover,  $\Gamma_{m,n} = A_{m,n} \exp(j\Phi_{m,n})$ , where  $A_{m,n}$  and  $\Phi_{m,n}$ 







**Figure 7.** Three-dimensional scattering patterns of the optimized coding metasurface for normal x-polarized electromagnetic incident waves at (a) 15 GHz, (b) 25.6 GHz, (c) 34.2 GHz, and (d) 45 GHz.

are the lattice reflection amplitude and phase coefficient, respectively, which are shown in Fig. 1(c,d) for "0" and "1" elements. We then define an array pattern function  $F(\theta, \varphi)$  for the  $M \times N$  lattice based on Eq. (1). It is usually safe to omit the effect of  $K_{m,n}$  and  $f_{m,n}(\theta, \varphi)$  for a reflective coding metasurface when calculating  $F(\theta, \varphi)$ . In this analysis, we only consider the phase difference between the "0" and "1" elements, and assume that  $A_{m,n}$  is fixed to simplify the calculation. Thus,  $F(\theta, \varphi)$  for the metasurface can be described as

$$F(\theta, \varphi) = \sum_{m=1}^{M} \sum_{n=1}^{N} \exp\left\{-j\left\{\Phi_{m,n} + k_0 D\sin\theta\left[\left(m - \frac{1}{2}\right)\cos\varphi + \left(n - \frac{1}{2}\right)\sin\varphi\right]\right\}\right\}.$$
(2)

Then, we employ a GA to configure an optimal element arrangement for the random coding metasurface. A flowchart of the GA applied in the present study is given in Fig. 3. The random initialization of a batch of feasible solution sets  $(N_p)$  comprising  $100 M \times N$ , matrices, where M = N = 12, which consist of an arbitrary



**Figure 8.** Scattering patterns of the optimized coding metasurface in the XOZ-plane at (a) 15 GHz, (b) 25.6 GHz, (c) 34.2 GHz, and (d) 45 GHz. Scattering patterns of the metal surface in the XOZ-plane at (e) 15 GHz, (f) 25.6 GHz, (g) 34.2 GHz, and (h) 45 GHz.

number of "0" and "1" elements distributed at random positions, represents the first generation of the search population. Equation (2) is employed as the evaluation function, and a threshold value of  $-10 \, dB$  is set to obtain an optimally high RCS reduction. Then, we establish uniform sampling points based on the values of  $\theta$  and  $\varphi$  according to 1° intervals to simplify the calculation, and generate 100 matrices with dimensions  $D_{\theta} \times D_{\varphi}$ , where  $D_{\theta}=91$  and  $D_{\varphi}=361$ . Subsequently, we compute the sequence of matrices according to the value obtained from Eq. (2) from small to large. If the result meets the requirements, it is recorded. To facilitate the calculation, we employ fixed values of the crossover and mutation probabilities  $p_c = 0.9$  and  $p_m = 0.1$ , respectively. In our GA, we employ single-point crossover as the most common crossover operator and binary mutation as the mutation operator. Then, the optimal solution is discovered by conducting 1000 iterative computations of the GA. After each iteration, the results are updated. After completion, we obtain the arrangement of "0" and "1" elements providing the minimum value of Eq. (2), which represents the metasurface providing the greatest RCS reduction. The normalized field intensity patterns obtained for coding metasurfaces with various uniform coding sequences lead to either a strong reflection beam or lobes projected toward the normal direction, as shown in Fig. 4(a-c). Comparing these figures with Fig. 4(d) obtained for the metasurface optimized according to the proposed procedure indicates that the optimized solution provides an ideal scattering reflection pattern.

**Simulations and measurement results.** Simulations of the optimal coding metasurface obtained by GA shown in Fig. 5 were conducted using CST Microwave Studio. For comparison, identical simulations were conducted with a metal (PEC) surface of equivalent geometry to that of the metasurface. When both *x*- and *y-polarized* waves are normally incident on the metal surface and metasurface, the RCS appears as given in Fig. 6(a). Figure 6(b) shows a broadband 10 dB reduction in the RCS of the metasurface over frequency ranges from 17 GHz to 42 GHz. The RCS reduction between 24 GHz and 26 GHz is slightly more than 20 dB. A broadband RCS reduction more than 15 dB is observed for a 4 GHz bandwidth from 23.1 GHz to 27.1 GHz, and the RCS reduction presents a maximum of 25 dB at 25.6 GHz under y-polarized electromagnetic incident waves. Figure 7(a–d) show the simulated three-dimensional (3D) scattering patterns of the metasurface at 15 GHz, 25.6 GHz, 34.2 GHz, and 45 GHz, respectively. The radiation patterns shown in the Fig. 4(d) are in agreement with the full-wave simulations at 25.6 GHz. To compare the scattering properties of the metasurface and the



**Figure 9.** RCS of the metal and optimized coding metasurface under obliquely incident electromagnetic waves from 10° to 40°: (**a**) bistatic RCS; (**b**) RCS reduction.



**Figure 10.** The experimentally obtained RCS (**a**) and RCS reduction (**b**) for various incident angles with vertical polarization.

metal surface quantitatively, the scattering patterns in the *xz-plane* obtained along a line passing through the middle of the surfaces (i.e., y = 0 [the *XOZ-plane*]) were obtained by simulation at 15 GHz, 25.6 GHz, 34.2 GHz, and 45 GHz, and the patterns for the metasurface and the metal surface are shown in Fig. 8(a–h), respectively. According to the law of energy conservation, the RCS reduced metasurface must have a repressed main lobe and enforced side lobes to maintain the overall scattering energy. Indeed, it can be observed in Fig. 8(e–h) that a strong main lobe appears in the backward direction for the metal surface in the full working band. In Fig. 8(b,c,f,g), the metasurface is observed to have suppressed the main lobe energy, and uniformly scattered energy to the side lobes at 25.6 GHz and 34.2 GHz relative to the metal surface. To investigate the scattering properties of the metasurface under oblique incidences, electromagnetic waves impinging from 10°, 20°, 30°, and 40° were considered. Due to the properties of the unit cells, linear polarization conversion can be realized over a wide range of oblique incidence angles. The RCS and RCS reduction over a broadband at the incidence angles considered can be observed in Fig. 9(a,b), respectively.

To experimentally verify the performance of the optimal random coding metasurface, a sample was fabricated and measurements were conducted, as described in the Method section. In addition, a metal sheet of equivalent geometry was investigated as a reference. Figure 10 provides the experimental data for vertically polarized incidence of varying angles. Note the test environment is not a completely anechoic chamber and the metal is lossy, energy of the high-frequency reflected electromagnetic waves decrease sharply. In addition, the dielectric constant of F4B become increasingly unstable with the increase of frequency, leading to the shifts of the resonance frequency. However, both the metal and metasurface RCS results have the same drop rate. Thus, the metasurface can be considered reasonably to hold the characteristic of RCS reduction. The figure indicates that RCS reduction of over 10 dB is generated in the frequency range from 17 GHz to 40 GHz, and comparison between Figs 9(b) and 10(b) demonstrate generally good agreement between the simulation and experimental results in terms of the overall trends. While the metasurface unit cell exhibits phase aberrations at various angles of incidence, the RCS reduction decreases only slightly. However, overall, the proposed metasurface performs well.

#### Discussion

In this paper, a broadband and broad-angle polarization-independent random coding metasurface for RCS reduction has been proposed and verified through simulations and experiments. Based on diffuse reflection theory and the scattering pattern reshaping method, a GA was introduced to obtain the optimal coding matrix composed of arbitrarily placed "0" and "1" elements. Excellent agreement between the simulated and the measured results demonstrate that the optimal random coding metasurface can efficiently realize broadband RCS



**Figure 11. (a)** The fabricated 1-bit GA-optimized coding metasurface. **(b)** The experimental setup for verifying RCS reduction.

reduction more than 10 dB from 17 GHz to 42 GHz when the angle of incident waves varies from 10° to 50°. The characteristic of broad bandwidth and broad-angle of incident waves for RCS reduction makes it promising for electromagnetic cloaking in microwave regime.

### Methods

The metasurface sample (shown in Fig. 11(a)) was placed at an equivalent height as the antennas in experiment, and the distance between sample and antennas was sufficient to avoid near field effects. The measurement setup is shown in Fig. 11(b). Pyramidal absorbing materials were placed around the sample to decrease background noise. In addition, a metal sheet of equivalent geometry was measured as a reference standard. To obtain the extent of RCS reduction, both the scattering coefficients of the metal and of the sample were measured. Two sets of ridged horn antennas were employed as transmitter and receiver, where both included frequency ranges of 15 GHz–18 GHz, 18 GHz–26.5 GHz, and 26.5 GHz–40 GHz. The antennas were connected to a vector network analyzer, which has the function of time domain gating. The receiver antenna can be reconfigured between transverse-electric and transverse-magnetic modes when rotated by 90°. The test bench allowed a 360° rotation to enable measurement of the RCS at various scattering angles.

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#### **Author Contributions**

H.Y.S., C.Q.G. and X.L.C. conceived the idea. H.Y.S. designed the sample and did the simulation. H.Y.S. and X.L.C. performed the experiment. Z.L., L.L.L., B.Z.X. and Z.C.Z. helped with the experimental measurements. H.Y.S., C.Q.G. and Z.L. wrote the manuscript based on the input from all authors. All authors contributed to the discussion.

#### Additional Information

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