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Full Cycle Rapid Scan EPR Deconvolution Algorithm

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Abstract

Rapid scan electron paramagnetic resonance (RS EPR) is a continuous-wave (CW) method that combines narrowband excitation and broadband detection. Sinusoidal magnetic field scans that span the entire EPR spectrum cause electron spin excitations twice during the scan period. Periodic transient RS signals are digitized and time-averaged. Deconvolution of absorption spectrum from the measured full-cycle signal is an ill-posed problem that does not have a stable solution because the magnetic field passes the same EPR line twice per sinusoidal scan during upand down-field passages. As a result, RS signals consist of two contributions that need to be separated and postprocessed individually. Deconvolution of either of the contributions is a wellposed problem that has a stable solution. The current version of the RS EPR algorithm solves the separation problem by cutting the full-scan signal into two half-period pieces. This imposes a constraint on the experiment; the EPR signal must completely decay by the end of each half-scan in order to not be truncated. The constraint limits the maximum scan frequency and, therefore, the RS signal-to-noise gain. Faster scans permit the use of higher excitation powers without saturating the spin system, translating into a higher EPR sensitivity. A stable, full-scan algorithm is described in this paper that does not require truncation of the periodic response. This algorithm utilizes the additive property of linear systems: the response to a sum of two inputs is equal the sum of responses to each of the inputs separately. Based on this property, the mathematical model for CW RS EPR can be replaced by that of a sum of two independent full-cycle pulsed field-modulated experiments. In each of these experiments, the excitation power equals to zero during either up- or down-field scan. The full-cycle algorithm permits approaching the upper theoretical scan frequency limit; the transient spin system response must decay within the scan period. Separation of the interfering up- and down-field scan responses remains a challenge for reaching the full potential of this new method. For this reason, only a factor of two increase in the scan rate was achieved, in comparison with the standard half-scan RS EPR algorithm. It is important for practical use that faster scans not necessarily increase the signal bandwidth because acceleration of the Larmor frequency driven by the changing magnetic field changes its sign after passing the

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inflection points on the scan. The half-scan and full-scan algorithms are compared using a LiNC-BuO spin probe of known line-shape, demonstrating that the new method produces stable solutions when RS signals do not completely decay by the end of each half-scan.

Graphical abstract



1. Introduction

Conventional continuous-wave (CW) spectrometers, which operate on principles developed in the last century, remain major electron paramagnetic resonance (EPR) instruments used worldwide. Magnetic field modulation with an amplitude smaller than the spectral linewidth is used to measure first-derivative spectra. The standard CW experiment is not fully optimized for EPR sensitivity but rather for simplicity of data acquisition and processing, which was an important requirement in the pre-digital era. Only a small fraction of the spins contributes to EPR signal and the resonator bandwidth is often much larger compared to the signal bandwidth. Commercially available fast digitizers and computers have made it possible to reinvent the CW method. The amplitude of the sinusoidal modulation can be increased by one to two orders of magnitude to encompass the entire EPR spectrum. All paramagnetic centers in the sample would contribute to the EPR signal twice during the modulation period, resulting in a signal gain [1]. Because the spin system is not readily saturated in the rapid passage regime, a higher excitation power than in the standard experiment can be used without causing spectral distortions. Improved sensitivity comes with a data processing challenge. In comparison with the standard first-harmonic CW method, periodic response of the spin system to a full-spectrum sweep is a Fourier series of tens to hundreds high-order harmonics of the modulation frequency. Fast magnetic field scans generate transient signals, analogous to free induction decay (FID) responses to an excitation pulse [2]. The mathematical problem of signal-spectrum transformation was solved in the last century for nuclear magnetic resonance [3, 4] and recently for EPR [1] for the special case of a linear scan. The development of an algorithm that permits the use of a sinusoidal magnetic field stimulus was a turning point in the evolution of CW EPR because producing a single harmonic current is much less of an engineering challenge compared to generation of a saw-tooth waveform in a coil [5–7]. The coil inductance can be compensated with a capacitor to make a resonance circuit for a specific frequency. Fast scans that span

over wide EPR spectra can be produced [8]. The rapid scan (RS) EPR method of generating and processing transient responses to the sinusoidal magnetic field stimulus was quite a success, producing a multitude of publications [9–25], a patent application [26], and commercialization by Bruker BioSpin [27]. The current version of the RS EPR algorithm solves an ill-posed problem – that the solution is unstable to small variations in the experimental parameters and noise - as two independent well-posed problems. Full-cycle RS signals are divided into two halves for up- and down-field scans [5], and each component is deconvolved separately. The half-scan RS EPR algorithm imposes a restriction on the experiment: transverse magnetization must decay completely by the end of each half-scan to avoid truncation of the signals. This limits the scan frequency and, therefore, the sensitivity. A full scan deconvolution algorithm is proposed in this paper that does not require signal truncation because it utilizes the additive property of linear time-invariant systems, as described in detail in the Theory section of this paper. The new full-cycle algorithm still requires a decoupling of up- and down-field scan contributions to the RS signal. The use of the previously developed background removal procedure [28] permits frequency domain separation of the contributions. The method does not work if up- and down-scan signals overlap. As a result, only a two-fold increase in the scan rate can be achieved, in comparison with the half-scan RS EPR deconvolution algorithm. This limitation will be resolved by the development of new signal separation methods.

2. Theory

Experimentally measured RS signal is an averaged, amplified, and down-converted response of the spin system m(t):

 $m(t)=m_x(t)+jm_y(t),$ (1)

to a stimulus in the form of a sinusoidal magnetic field

$$B(t) = -0.5 B_{pp} \cos(2\pi f_s t)$$
 (2)

in the presence of CW excitation. B_{pp} is the peak-to-peak amplitude, and f_s is the frequency of the alternating field. Because the periodic stimulation B(t) produces a periodic response, then

$$m(t)=m(t+P), P=1/f_s,$$
 (3)

and measurement of transient EPR coherences that last longer than P can be problematic.

2.1 Linear system approximation

The rapid scan data processing algorithm assumes that the spin system is well-approximated as a linear time-invariant system (LTI). Any LTI system is uniquely characterized by the impulse response function, which is an FID in the case of magnetic resonance. FID is an

outcome of an excitation in the form of an infinitely narrow pulse. Within the LTI assumption, an arbitrary pulse, $B_1(t)$, produces response m(t):

$$m(t)=B_1(t)\otimes FID(t),$$
 (4)

where \otimes is the convolution operator. If an input $B_1^a(t)$ produces an output $m^a(t)$ and an input $B_1^b(t)$ generates an output $m^b(t)$, then the sum of the inputs gives the sum of the outputs:

$$(B_1^a + B_1^b) \otimes FID = B_1^a \otimes FID + B_1^b \otimes FID = m^a + m^b = m \quad (5)$$

Dependence on time in Eq. (5) is omitted for better readability. An important point of the full scan deconvolution algorithm described below is that Eq. (5) can also be used in reverse. An actual excitation field, $B_1(t)$, can be represented as a sum of two arbitrary functions, and the observed in the experiment spin response can also be separated into two contributions corresponding to each of the inputs.

Using the well-known property of Fourier transformation (FT), convolution in Eq. (4) can be changed into multiplication in the frequency domain:

$$FT[m(t)] = FT[B_1(t)] FT[FID(t)],$$
 (6)

EPR spectrum $S(\omega)$ can be found from Eq. (6):

$$S(\omega) = FT[FID(t)] = FT[m(t)] / FT[B_1(t)]$$
 (7)

Because both $B_1(t)$ and m(t) functions are periodic, FID(t) is also periodic with the same period P. As a result, any spin response signal that lasts longer than P cannot be measured without truncation:

$$1/f_{s} > P_{min} = N T_{2}^{*}$$
 (8)

Eq. (8), in which T_2^* is an observed transverse relaxation time, establishes the upper theoretical limit for the scan frequency f_s . The factor N in Eq. (8) can be in the range of three to five, depending on the accepted linewidth broadening. For certain types of experiments lineshape accuracy is not as important as signal intensity.

The spin system is well-approximated as LTI in the conventional CW experiment if the excitation power is below the saturation level. In the rapid scan regime, Eqs. (4–7) are valid at higher compared to the standard method powers. The faster the scan, the larger B_1 values can be used without causing spectral distortions. In the extreme case of the scan frequency approaching the limit given by Eq. (8), the RS signal resembles that of an FID in the pulsed EPR experiment. The spins experience a short-duration excitation as B(t) passes through the

resonance. Free evolution of the spins in the changing B(t) field is observed after the excitation. In the system of coordinates associated with B(t)-driven Larmor frequency, the spins experience a short frequency modulated pulse (chirp pulse).

2.2 Transformation into reference frame associated with the Larmor frequency

RS EPR is a CW experiment, i.e., the excitation field is time-independent in the frame associated with the excitation source frequency. Direct substitution of a constant B_1 value in Eqs. (4–7) does not produce any meaningful results. In addition, there is also no explicit formulation of the scanning field B(t) within the LTI mathematical model. The solution to this problem lies in transformation of the RS signal into a reference frame associated with the Larmor frequency of the spins $\omega_L(t)$:

$$\omega_{\rm L}(t) = \gamma B(t) = -0.5 \gamma B_{\rm pp} \cos \left(2\pi f_{\rm s} t\right).$$
(9)

To obtain a mathematical expression for the transformation, one should find the phase difference between the laboratory and accelerating frames. The phase gain starting from the beginning of the scan (t = 0) can be found by integration of Eq. (9):

$$\varphi(t) = \int_0^t \gamma B(\tau) d\tau = -\frac{\gamma B_{pp}}{4\pi f_s} \sin(2\pi f_s t).$$
(10)

In the frame associated with $\varphi(t)$, B(t)=0. The transformation into the system of coordinates associated with $\omega_L(t)$ can be achieved by multiplication:

$$\mu(t) = m(t)d(t), d(t) = \exp\left[-i\varphi(t)\right]$$
 (11)

For an observer in the new system of coordinates, the laboratory frame is accelerating, and the B_1 field becomes time-dependent:

$$B_1(t)' = B_1 d(t).$$
 (12)

Applying Eqs. (11–12) is equivalent to a change of the experiment; field-swept CW is replaced by frequency-swept EPR at a constant external field. The transformation permits Eq. (4) to be rewritten in the form:

$$\mu(t) = B_1 d(t) \otimes FID(t).$$
 (13)

Deconvolution of EPR spectrum from RS signals can be done as follows:

$$S(-B)=S(\omega/\gamma)\propto FT[\mu(t)]/FT[d(t)]$$
 (14)

The minus sign in front of B in Eq. (14) reflects the fact that field and frequency scans occur in opposite directions. In the up-field scan CW experiment, the spins at a higher local field are observed first. An up-frequency scan at the constant external field excites these spins last.

2.3 Half-cycle deconvolution

Eq. (14) is a classic example of deconvolution, which is often an ill-posed problem. A unique solution may not exist or/and be unstable, which is the case for RS EPR. The absolute value of the denominator in Eq. (14) is an oscillating function with multiple values approaching zero [5], as shown in Fig. 1c

It happens because B(t) in Fig. 1a passes through the same value of the magnetic field twice during the period (in Fig. 1 B(t)=-1 is chosen as an example). The frequency with which d(t) oscillates (see Fig. 1b) is equal to $\omega_L(t)$. After FT, two contributions, corresponding to the same frequency but different time instances, are summed. If the contributions are inphase with each other, there is a constructive interference. If they are 180° out-of-phase, they interfere destructively and FT[d(t)] approaches zero, as shown in Fig. 1c. As a result, the deconvolution problem in Eq. (14) is ill-posed for two reasons. First, multiple near-zero values of the denominator amplify noise. More importantly, the positions of the zeros strongly depend on the parameters of the scan. Miniscule uncertainty in the estimation of either the experimental scan amplitude and/or phase results is instability of the solution. A stable solution of the deconvolution problem can be found by imposing a restriction on the experiment; transverse magnetization must completely decay by the end of the up- or downscan, as shown in Fig. 2. In this case, the RS EPR signal can be separated into two halves (see Fig. 2b), and each half-period piece can be deconvolved independently (see Eq. [14]):

 $S^{up}(-B) \propto FT[\mu(t)]/FT[d(t)], 0 \le t < P/2$ (15.a)

 $S^{dn}(-B) \propto FT[\mu(t+P/2)]/FT[d(t+P/2)], 0 \le t < P/2$ (15.b)

Because B(t) doesn't pass through the same field twice, the denominators do not oscillate as much (see Fig. 1d), and the deconvolution problem becomes well-posed.

It is important to note that Eq. (15a, b) describe experiments that are different from the actual one shown in Fig. 2a. These experiments correspond to a situation in which the magnetic field abruptly changes at the end of each half-scan, as shown in Fig. 2c. They are still CW experiments, but the scan periods are equal to P/2. As a result, the theoretical limit for the duration of FID is reduced by a factor of two (see Eq. [8]). As long as m(P/2)=m(P)=0, substitution of the actual full cycle experiment by a half-cycle experimental model is acceptable.

2.4 Full cycle deconvolution

An alternative approach to the deconvolution problem that is based on the property of linear systems described in Eq. (5) is suggested in this paper. It t does not require the restrictive change of the experimental model. A constant B_1 amplitude can be represented as a sum of two step functions for up- and down-scans:

$$B_1 = B_1^{up}(t) + B_1^{dn}(t),$$

$$B_1^{up}(t) = B_1 \theta(t) \theta(-t + P/2),$$
 (16)

$$B_1^{dn}(t) = B_1 \theta(t - P/2) \theta(-t + P),$$

$$\theta(t) = \begin{cases} 1, t > 0 \\ 0, t < 0 \end{cases}$$

shown as blue and red traces in Fig. 3b, respectively.

The exact and stable solution of the full scan deconvolution problem without any restrictions can be found by using Eqs. (5, 13, 14):

$$S^{up}(-B) \propto FT[\mu^{up}(t)]/FT[d(t)B_1^{up}(t)], 0 \le t < P$$
 (17.a)

$$S^{dn}(-B) \propto FT[\mu^{dn}(t)]/FT[d(t)B_1^{dn}(t)], 0 \le t < P$$
 (17.b)

It is because the denominators are well-defined functions. The challenge for the practical use of Eq. (17) is to separate experimental signal m(t) into two components, corresponding to $B_1^{up}(t)$ and $B_1^{dn}(t)$. The method that was developed for background removal in the half-cycle deconvolution algorithm can be used to accomplish this goal [28]. When a magnetic field passes through the resonance in the up-direction, the precession frequency of the spins increase. Passage in the down-direction is associated with a decrease of the Larmor frequency. As a result, the up-and down-scan EPR signals can be separated in the frequency domain. However, this method has a limitation, which can be demonstrated in the example shown in Fig. 1. After passing the resonance value of B(t) = -1 (first red circle), the precession frequency will continue to increase until it reaches the value of B(t)=2 at the scan extremum. After this point, the Larmor frequency slows down but is still larger than that at the resonance position until it comes to resonance again at B(t)=-1 during the second half-scan. This position is denoted as a red circle as well. The down-scan signal is excited at this

point, which can overlap with any residual of the up-scan signal both in the time and frequency domains. The FT-based separation method cannot distinguish these two overlapping signals, imposing restriction of the scan frequency. The RS signal must completely decay between the two re-excitations (between the two red cycles in Fig. 1). This constraint is not as limiting as that imposed by the half-scan algorithm; twice as fast scans can be used with the new algorithm. Further development of the algorithm will permit separation of the overlapping up- and down-field signals that can be achieved, for example, by varying the scan parameters.

Scan rate can also be limited by the bandwidth of the detection resonator, which acts as a band-pass filter for RS signals. In the example of Fig. 1, the transverse magnetization generated at the resonance position B(t)=-1 G will precess with increasing frequency until it reaches the highest point on the scan, B(t)=2 G. At this point, the Larmor frequency will be approximately 9 MHz higher than the excitation frequency. The resonator bandwidth must be large enough to permit an undistorted detection of a 9 MHz component of the signal. As explained above, after passing the maximum field point on the scan, the precession frequency decreases, and the signal bandwidth remains the same. Thus, increasing the frequency for full-cycle deconvolution would not necessarily require reduction of the resonator quality factor, which may be detrimental to EPR sensitivity.

2.5 Sampling interval in the spectral dimension

Any periodic function has a discrete spectrum defined at the integer multipliers of the fundamental frequency. For this reason, FT of m(t) and d(t) are sampled with the interval equal to $d\omega = 2\pi f_s$ in the frequency domain. In the magnetic field domain, EPR spectra are discrete, with the sampling increment equal to:

$$\Delta B = 2\pi f_s / \gamma \quad (18)$$

The number of points in the deconvolved spectrum is equal to

$$Np=B_{pp}/\Delta B$$
 (19)

Np is rounded to the closest integer number. Because m(t) and d(t) functions are truncated in the half-scan algorithm, the effecting scan frequency is twice as large so that the sampling interval is equal to 2 B and the number of data points reduces to Np/2. Thus, EPR spectra become under-sampled by a factor of two. This is not a problem if f_s is relatively small but may be detrimental for high frequency scans that approach the limit of Eq. (8).

Experiment

3.1 Sample

Locally synthesized Lithium octa-n-butoxy 2,3-naphthalocyanine (LiNc-BuO) microcrystals were used to compare the half-cycle and full-cycle deconvolution algorithms [29, 30]. LiNC-BuO crystals were placed into a 1 mm ID tube, degassed, and flame-sealed. The sample

height was approximately 2 mm. LiNC-BuO is used in EPR oximetry, including clinical applications as an oxygen sensitive core of the OxyChip [31]. The spin probe is known to have a Lorentzian line shape, width of which is oxygen dependent.

3.2 Spectrometer

A home-built RS EPR spectrometer/imager was used for the experiments. Even though not all intended pieces of the design have been implemented, the spectrometer is fully functional. Its performance was tested using a variety of samples with the known EPR spectra. A detailed description of the final design will be published elsewhere.

A permanent magnet (Ningbo Jansen NMR Technology Co) was equipped with sweep coils to produce a magnetic field in the range 268 ± 25 G, which corresponds to a frequency span from 680 to 820 MHz. A single loop reflection resonator was manufactured using a previously described design for 300 MHz EPR imaging of rats [32]. The sample was placed into a tubed filled with salted water. The loaded quality factor, O, was approximately 100 at a 767 MHz operating frequency. Rapid scan coils were locally built using Litz wire wound on a 3D printed plastic support structure that also accommodated a mouse bed, which was not used in the experiments described here. Sinusoidal scans were produced by a Keysight 33622A arbitrary waveform generator (AWG). The AWG output was amplified using an 1800 watt Cerwin-Vega CV-1800 audio amplifier. The RS coils that had Helmholtz configuration were resonated using a series of capacitors. The capacitors were inserted into three breadboards. Two boards were connected to each coil of the Helmholtz pair and the third board was used to insert a current probe capacitor. This design permitted selection of the desired frequencies, which were 29, 72, and 98 kHz in the experiments described below. The phase and amplitude of the current in the coils were measured using a Teledyne AP031 differential probe that was connected to the two terminals of a 100 nF probe capacitor. The waveforms were digitized using an NI PCIe 6363 DAQ card. Measured values were used for deconvolution.. The RF bridge, which was designed and built locally, output a baseband RS signal that was digitized using a Keysight U1084A high-speed digitizer. The card and external field were controlled by SpecMan4EPR software. The digitizer was triggered by the AWG RS signals were post-processed into absorption EPR spectra using locally-written Matlab programs.

3.3 Comparison of half-cycle and full-cycle deconvolution methods

RS data were collected at three scan frequencies: 29, 72, and 98 kHz. The signals were processed using the published half-cycle RS EPR algorithm and the new full-cycle deconvolution method. EPR spectra were fitted with a Lorentzian line-shape function using the Matlab Curve Fitting Tool (CFT). The CFT output the fitting results together with the confidence bounds at a level of 95% and root mean square errors (RMSE) of the fitting. The results are summarized in Table 1.

At the lowest frequency of 29 kHz, the scan rate is relatively slow, and RS signals decay completely by the end of each half-scan. Both algorithms give similar half-magnitude (FWHM) linewidth estimations of 123 mG within narrow confidence intervals and low root mean square error (RMSE). The measured linewidth of the deoxygenated sample was

confirmed by an independent measurement using a Bruker X-band EPR spectrometer. At higher scan frequencies, both algorithms show increases in FWHM. The fitting results of the spectra deconvolved using the half-cycle method demonstrate larger linewidth broadening, increased confidence intervals, and RSME. The uncertainty caused by under-sampling (see Eqs. 18, 19) can be partially reduced by interpolation using the Matlab routine interpft. Fig. 4 demonstrates EPR spectra obtained using the two algorithms before (Fig. 4a, c) and after (Fig. 4b, d) interpolation. EPR spectra in Fig. 4a, b demonstrate oscillations that is the result of RS signal truncation.

4. Results and Discussion

A sinusoidal magnetic field scan, B(t), passes through the same EPR line twice during the scan period. Periodic RS signal is a response to two consequent spin system excitations. Recovery of the EPR spectrum from a transient signal is an ill-posed problem. However, in linear system approximations, periodic m(t) can be represented as a sum of the responses to two excitations that occur during up- and down-field scans. Based on this assumption, a half-scan deconvolution algorithm was developed that requires complete decay of the transverse magnetization by the end of each half-scan (see Fig. 2). The RS signal m(t) is split into two equal pieces for $0 \le 1/2$ and $P/2 \le 1/2$ that are post-processed individually. The algorithm imposes an upper limit of the scan frequency and may generate under-sampled EPR spectra, as shown in Fig. 4 a. The full-scan deconvolution algorithm described in this paper does not require truncation of the signal. Provided that the responses for up-field and down-field scans can be separated from m(t), the algorithm outputs stable solutions with no restrictions beyond those imposed by the spin system itself. The restrictions include a requirement for linearity (see Eq. 5) and the relaxation time limit expressed in Eq. (8). Previously developed background removal was used for the separation of the up- and downscan contributions. This method can be used if there is no overlap between $m_{un}(t)$ and $m_{dn}(t)$ contributions, as shown in Fig. (3). As a result, the current version of the full-scan algorithm permits only a two-fold increase of the scan frequency, compared to the half-scan method. Further developments will make it possible to reach the theoretical limit given by Eq. (8). It is important to note that a scan rate increase does not necessarily affect the signal bandwidth. RS EPR is a narrowband excitation and broadband detection method that is in comparison with the standard CW (pure narrowbanded) and pulsed (pure broadbanded) methods. The highest signal-to-noise ratio can be achieved when the signal bandwidth matches that of the detection resonator. Very small spectral broadening in the data presented in Table 1 is the result of the resonator bandwidth being slightly smaller than the signal bandwidth. No statistically significant line broadening was observed between the results obtained at 72 and 98 kHz. This is because the Larmor frequency reaches its maximum (or minimum) at the inflection points of the scan. The new full-cycle deconvolution algorithm permits a two-fold increase in the scan frequency without increasing the signal bandwidth. As a result, twice as many signal averages per unit time can be performed to additionally improve signal-tonoise. The goal of this paper is to introduce a second-generation RS EPR algorithm and demonstrate its implementation. The use of this algorithm for EPR spectroscopy and imaging using various samples will provide more insight into the advantages of this new deconvolution method and ways to improve its performance.

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Highlights

Deconvolution of full cycle RS signal is an ill-posed problem

Separation of up- and down-field scan signals makes it well-posed

Currently used algorithm divides RS signal to two halves

Half-cycle deconvolution imposes restrictions on the scan rate

New RS EPR algorithm deconvolves two full-cycle signal components

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Fig. 1.

Graphical explanation of the deconvolution problem ill-posedness. (a) Magnetic field passes through a resonance position (-1 G in the example) twice during the scan period in two separate instances of time. (b) Function d(t) oscillates with the Larmor frequency $-\gamma$ (ca. -2.8 MHz) at these moments. (c) Fourier transformation of d(t) sums two complex contributions corresponding to the same frequency to produce an interference pattern. Depending on the scan frequency and amplitude, the interference can be either constructive or destructive. As a result, FT[d(t)] in the denominator of Eq. (14) causes instability of the deconvolution solution. (d) There is no interference for half-scan d(t), and the solution of Eq. (14) is well-posed.



Fig. 2.

Changing the scan model from a full-cycle scan (a) to a half-scan (c) helps to find stable solutions for the deconvolution problem. If the RS signal completely decays before it reaches the inflection points of B(t), up- and down-scan contributions can be separated (b). $B_{up}(t)$ and $B_{down}(t)$ scans (c) are periodic, as well as B(t). The period is equal to P/2. In both models, the excitation field B₁ is CW.

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Fig. 3.

A full-scan solution of Eq. (14) that does not limit scan rates. RS signal may not decay by the end of a half-scan (a). Constant B_1 can be represented as a sum of up-scan (green) and down-scan (magenta) step-functions (b). Separation of m(t) in (a) into its corresponding excitations responses (c) and (d) permits the stable deconvolution described by Eq. (17).



Fig. 4.

EPR spectra (discrete points) obtained using half-cycle (a) and full-cycle (c) RS deconvolution methods. The sampling rate for (c) is twice as large. Fourier interpolation increases the sampling rate for (a) and (c) to give (b) and (d), respectively. Solid blue lines are the fitting functions of Lorentzian shape. Scan frequency and scan rate were 98 kHz and 920 kG/s, respectively.

Table 1

Comparison of EPR line fitting results for half-cycle and full-cycle RS deconvolution algorithms. EPR spectra obtained at 98 kHz were interpolated.

| | Half-cycle algorithm | | Full-cycle algorithm | |
|-----------------------|----------------------|-----------|----------------------|----------|
| Scan frequency | FWHM [mG] | RMSE *1e4 | FWHM [mG] | RMSE*1e4 |
| 29 kHz | 123±2 | 62 | 123±1 | 46 |
| 72 kHz | 131±5 | 133 | 127±1 | 51 |
| 98 kHz | 134±9 | 182 | 130±2 | 74 |
| 98 kHz (interpolated) | 135±4 | 160 | 129±1 | 65 |