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Diffuse correlation spectroscopy beyond the water peak enabled by cross-correlation of the signals from InGaAs/InP single photon detectors

Mitchell B. Robinson,

Harvard-MIT Program in Health Sciences and Technology, Massachusetts Institute of Technology, Cambridge, MA 02139 USA

Marco Renna,

Athinoula A. Martinos Center for Biomedical Imaging, Department of Radiology, Massachusetts General Hospital and Harvard Medical School.

Nisan N. Ozana,

Athinoula A. Martinos Center for Biomedical Imaging, Department of Radiology, Massachusetts General Hospital and Harvard Medical School.

Adriano Peruch,

Athinoula A. Martinos Center for Biomedical Imaging, Department of Radiology, Massachusetts General Hospital and Harvard Medical School.

Sava Sakadžić ,

Athinoula A. Martinos Center for Biomedical Imaging, Department of Radiology, Massachusetts General Hospital and Harvard Medical School.

Megan L. Blackwell,

Massachusetts Institute of Technology Lincoln Laboratory.

Jonathan M. Richardson,

Massachusetts Institute of Technology Lincoln Laboratory.

Brian F. Aull,

Massachusetts Institute of Technology Lincoln Laboratory.

Stefan A. Carp,

Athinoula A. Martinos Center for Biomedical Imaging, Department of Radiology, Massachusetts General Hospital and Harvard Medical School.

Maria Angela Franceschini

Athinoula A. Martinos Center for Biomedical Imaging, Department of Radiology, Massachusetts General Hospital and Harvard Medical School.

Abstract

Objective: Diffuse correlation spectroscopy (DCS) is an optical technique that allows for the non-invasive measurement of blood flow. Recent work has shown that utilizing longer wavelengths beyond the traditional NIR range provides a significant improvement to signal-to-noise ratio (SNR). However, current detectors both sensitive to longer wavelengths and suitable for clinical applications (InGaAs/InP SPADs) suffer from suboptimal afterpulsing and dark noise characteristics. To overcome these barriers, we introduce a cross correlation method to more accurately recover blood flow information using InGaAs/InP SPADs.

Methods: Two InGaAs/InP SPAD detectors were used for during *in vitro* and *in vivo* DCS measurements. Cross correlation of the photon streams from each detector was performed to calculate the correlation function. Detector operating parameters were varied to determine parameters which maximized measurement SNR. State-space modeling was performed to determine the detector characteristics at each operating point.

Results: Evaluation of detector characteristics was performed across the range of operating conditions. Modeling the effects of the detector noise on the correlation function provided a method to correct the distortion of the correlation curve, yielding accurate recovery of flow information as confirmed by a reference detector.

Conclusion: Through a combination of cross-correlation of the signals from two detectors, model-based characterization of detector response, and optimization of detector operating parameters, the method allows for the accurate estimation of the true blood flow index.

Significance: This work presents a method by which DCS can be performed at longer NIR wavelengths with existing detector technology, taking advantage of the increased SNR.

Keywords

Fluid flow measurement; Infrared detectors Biomedical applications of optical radiation; Semiconductor device modeling

I. Introduction

Diffuse correlation spectroscopy (DCS) is an established optical technique that allows for the non-invasive measurement of tissue blood flow by analyzing the temporal correlation of the intensity of multiply scattered light [1]. DCS has shown to be effective in monitoring blood flow in many clinical and research environments, including monitoring of cerebral blood flow at the bedside [2]–[4] and characterization of breast cancer lesions as malignant or benign [5], [6]. To improve sensitivity to the target tissue; which is typically below dermal layers, layers of subcutaneous fat, and bone, depending on the measurement location; larger source detector separations are used, reducing measurement SNR. Several approaches building on the traditional DCS technique have been proposed to address this challenge, including heterodyne detection [7]–[10], time resolved detection [11]–[14], ultrasound tagging of light [15], [16], and multi-speckle detection [17], [18]. In addition, recent developments in the literature [19]–[22] have shown the effectiveness of moving to wavelengths beyond the traditional NIR range (650 – 850 nm) due to reduced optical attenuation relative to the traditional NIR range, increased safe exposure limits as defined by the ANSI standard, an increased number of photons per unit energy, and a slower autocorrelation decay due to lower scattering at the longer wavelength. Based on previous

calculations and *in vivo* results [19], [22], approximately 10 times more photons will find their way to the detectors using 1064 nm wavelength compared to the traditional NIR range, allowing for the use of longer source-detector separations to improve sensitivity to deeper blood flows while maintaining a similar signal-to-noise ratio (SNR), with an added benefit of measuring a slower decorrelation rate, improving the fitting of the curve. These benefits are difficult to fully realize in practice due to shortcomings in the characteristics of detectors sensitive to light at 1064 nm. Single photon detectors at 1064 nm include superconducting nanowire single-photon detectors (SNSPD) and InGaAs/InP single photon avalanche diodes (SPADs). SNSPDs have superior detection characteristics (e.g., high photon detection efficiency, low dark count rate, no afterpulsing) but their low portability and high cost limit their suitability for clinical applications. InGaAs/InP SPADs, while portable and suitable for a clinical environment, have poor noise characteristics, including relatively high dark count rate (DCR) and excessive afterpulsing [23], [24]. An afterpulse is generated when electrons from a previous avalanche become trapped in the detector material and subsequently trigger another avalanche when the detector is rebiased. Advances in detector hardware provide future solutions to reduce afterpulsing [25], though current strategies to mitigate their effects include using detector hold off times in the range of tens of microseconds to allow trapped charges to diffuse out [26]. This strategy presents a challenge for DCS, where the long hold off time ($>10\mu\text{s}$) can obscure the signal fluctuations that occur on the order of several microseconds. To move closer to a DCS setup for measurements at 1064 nm that is suitable for clinical applications, we propose to use cross-correlation of optical signals from two InGaAs/InP SPAD detectors to correct for the shortcomings of the InGaAs/InP SPADs. The use of cross-correlation, taking advantage of the Hanbury Brown Twiss effect [27], reduces the artifacts in the intensity autocorrelation function by allowing for photon interarrival times shorter than the hold off time and removing the relationship between an afterpulse and its inciting count, though the presence of the afterpulsing counts in the detector streams still represents a source of correlated noise. Afterpulsing counts can extend the correlation present in the intensity signal to longer delay times, leading to an artificial slowing of the autocorrelation decay. In this work we show that this disruption can be accounted for, both through the optimization of the operating parameters of the detectors (temperature, excess bias voltage, and hold-off time) as well as modeling the effects of these afterpulse trains on the correlation function. By addressing these challenges, DCS systems using sources beyond the traditional NIR range can be built with suitable portability for many applications.

II. Materials and methods

A. Diffuse correlation spectroscopy

Diffuse correlation spectroscopy relates the observed temporal fluctuations of light intensity to the motion of optical scatterers in the tissue. The rate of decay of the temporal intensity autocorrelation function, $g_2(\tau)$, is first related to the temporal electric field autocorrelation function, $g_1(\tau)$, by the Siegert relation [28], given as

$$g_2(\tau) = 1 + \beta |g_1(\tau)|^2, \quad (1)$$

where β is the coherence factor of the measurement that is dependent upon the measurement geometry, number of modes collected, coherence of the laser source, and presence of noise counts. The electric field autocorrelation function can be calculated as a weighted sum of electric field autocorrelation functions along individual pathlengths, $g_{1,s}(\tau) = \exp\left(-\frac{1}{3}n^2k_0^2\langle\Delta r^2(\tau)\rangle\frac{s}{l^*}\right)$, over the pathlength distribution for the measurement given as,

$$g_1(\tau) = \int_0^\infty P(s)e^{-\frac{1}{3}n^2k_0^2\mu_s'\langle\Delta r^2(\tau)\rangle s} ds, \quad (2)$$

where n is the index of refraction of the sample, k_0 is the wavenumber of the light in vacuum, μ_s' is the reduced scattering coefficient, $\langle r^2(\tau) \rangle$ is the mean squared displacement at time delay τ , and s is the pathlength of the photon. Blood flow in the tissue is described by the blood flow index (BFi), which is related to the mean squared displacement of the scatterers (e.g., red blood cells) by the following expression, $\langle r^2(\tau) \rangle = \alpha 6D_b\tau = BF_i\tau$, where α is the fraction of scattering events that happen at moving scatterers and D_b is the apparent diffusion coefficient. In this study, data are fit using the correlation diffusion equation for a homogenous, semi-infinite slab. Though blood flow at a macroscopic level does not appear to be diffusive, shear-induced diffusion of red blood cells in blood vessels has been investigated as the dominant source of the diffusive nature of experimentally measured flows [29]–[31], though other flow models have been proposed to describe the measured flow dynamics [32], [33].

B. Hardware

We used a custom built DCS system consisting of a 1064 nm laser source, two InGaAs/InP SPADs (PDM-IR fiber-pigtailed, Micro Photon Devices), and a custom FPGA-based software correlator board, described previously[34]. For the continuous wave (CW) DCS experiments, a long coherence length laser at 1064 nm (RFLM-125-0-1064, NP Photonics) was used. For the gated measurements, a custom-built pulsed laser at 1064 nm was used, generating pulses with a full width half maximum of 250 ps at a repetition rate of 25 MHz. In both the CW and pulsed experiments, 100 mW of incident power over a spot size larger than 1 mm was used, conforming to the ANSI standard for the maximal permissible exposure for skin. Light from the sample was collected using a 6.2 μm diameter fiber (1060 HI) and sent to both detectors using a 50/50 fused fiber coupler (TW1064R5F2A, Thorlabs). A reference measurement was made using a superconducting nanowire detector (SNSPD) (Quantum Opus) tuned for maximum efficiency at 1064 nm (~85–90%) for comparison. For the CW experiments, the InGaAs/InP detectors were set to free running mode. For gated experiments, a gate width of 5 ns was used to capture the entire temporal point spread function (TPSF), leading to measurements of the DCS autocorrelation that are effectively CW. In all cases, reference detectors were run in a free running mode.

C. Detector characterization experiments

To optimize the operating parameters of the InGaAs/InP detectors, we characterized the detector performance in both CW and gated mode. An intralipid solution with a reduced

scattering coefficient (μ_s') of 6 cm^{-1} and an absorption coefficient (μ_a) of 0.14 cm^{-1} at 1064 nm was measured at a source-detector separation of 1.2 cm, while the detectors' temperature, excess bias voltage, and hold off time were varied. Detector settings were optimized based on the SNR of the plateau of $g_2(\tau)$ at short time lags ($SNR = \langle g_2(10^{-6} \text{ s to } 10^{-5} \text{ s}) - 1 \rangle / STD(g_2(10^{-6} \text{ s to } 10^{-5} \text{ s}))$). The intralipid sample was measured for 60 s at each detector setting, and autocorrelation functions were calculated at a rate of 0.1 Hz. The range of the operating parameters can be seen below in Table I. *In-vivo* experiments were performed using the gated mode only. Measurements of the forehead were done at a source-detector separation of 1 cm. An elastic band was placed over the probe, and pressure was applied to the probe by tightening the tension in the band. A baseline of 60s was first acquired, followed by 60s of tourniquet tightening and 60s of recovery. Autocorrelation functions were calculated at 0.25 Hz, and a 12s-long moving average was applied to remove residual cardiac pulsation present in the fitted BF_i . The protocol was repeated 3 times, and the results were averaged. This study was reviewed and approved by the Mass General Brigham Institutional Review Board (#2019P003074; approved February 05, 2021).

D. Detector modeling

As seen in the results section below, the use of cross-correlation successfully removes large artifacts related to the afterpulsing and hold off time, making these InGaAs/Inp detectors usable for DCS measurements. Still the presence of afterpulsing counts results in a distortion of the correlation function that needs to be addressed. To remove the remaining distortion of the correlation function (i.e. the slower decay) after the calculation of the cross-correlation, we model the operation of each individual SPAD in a state-space representation. By discretizing time into bins set by the clock frequency of the correlator board (150 MHz), we can describe the state of the detector at all times through the following model, seen below in Fig. 1.

The detector operates in two phases, inactive and active. The length of the inactive period, the hold off time, is set in terms of M clock cycles. Each state, S_i , represents the detector at a particular number of clock cycles after the detector has entered the hold off time. Following the activation of the detector at state S_{m+1} , the detector can detect a photon that is either a true count, a (thermal) dark count, or an afterpulse. Upon a detection event, the simulated detector is quenched during the state transition between the detection state and state S_1 , and the simulated detection electronics register that the photon has arrived during state S_1 . In this model, the probability of detecting a signal photon or registering a dark count event are lumped into the probability p . The probability of detecting an afterpulsing count is given for each active state as A_i . Based on previous literature, the probability of afterpulsing is modeled as a power law decay, given as $A_i = \frac{AP_0}{f_s} \left(\frac{i}{M+1} \right)^{-\alpha}$, where AP_0 is the effective afterpulsing count rate during the first active state, f_s is the sampling frequency used for the simulation, and α is the power coefficient [24]. To determine how many active states, N , need to be considered, we impose that $P \gg A_{m+n}$ such that an afterpulse would be extremely unlikely beyond that state in comparison to the signal photon or dark count detection.

We propose to use this model to understand both the steady state and dynamic performance of the detector to determine the detector characteristics (P , AP_0 , and α) governing the state

transition probabilities. With q_i being the probability of being in a particular state, S_i , for the detector in the steady state, we can derive the equations to describe the distribution of probability throughout the states. As we are dealing with probabilities, all should sum to 1, as given below in (3),

$$\sum_{i=1}^{M+N} q_i = 1 \quad (3)$$

To investigate the steady state response of the detector, we impose the assumption that the probability of entering a state S_i is equal to the probability of leaving the state. We can express the steady state probability for each state, q_i , in terms of the probability of the detector being in the previous state, q_{i-1} , and the associated state transition probabilities. For the inactive states, the state transition probability to go from one state to the next is equal to one, which gives the relation $q_1 = q_2 = \dots = q_{M+1}$. For the active states, we can define the probability of entering the state as the probability of being in the previous state multiplied by the probability of not detecting a photon in that state, while the probability of exiting the state is equal to one. Balancing the probability of entering and exiting the state gives the relation $q_i = q_{i-1}(1 - A_{i-1} - P)$. For the last state, S_{m+n} , the use of the recurrence relation gives a different probability of exiting the state. The probability of entering the state will be the same as previously described for other active states, but the probability of exiting the state is given as the probability of being in state S_{m+n} , q_{M+N} , multiplied by the probability of detecting a real photon or a dark count, P . Balancing these probabilities gives the following relation, $q_{M+N} = \frac{q_{M+N-1}}{P}(1 - A_{M+N-1} - P)$. The description of these steady state probabilities is summarized in (4). Here, we use the recurrence relation at state S_{m+n} to simplify the description of the model when the afterpulsing probability, A_{m+n} , is effectively equal to zero. If instead we used an infinite number of active states after S_{m+n} , we could continue to express the probability of being in the series of states as $q_{M+N+m} = R_{M+N+m-1}(1 - P)$, where m is the number of states past state S_{m+n} . If we then express the total probability of being in this series of states we would have the relation $q_{M+N} \sum_{i=0}^{\infty} (1 - P)^i$, which reduces to q_{M+N}/P . Then using the relation for active states with afterpulsing, we can calculate the probability of being in state S_{m+n} by taking the probability of being in state S_{m+n-1} multiplied by the state transition probability, $1 - A_{m+n-1} - P$. This description gives the same total probability for the states without afterpulsing as the model containing the recurrence relation.

$$q_i = \begin{cases} q_1, & 1 \leq i < M + 2 \\ q_{i-1}(1 - A_{i-1} - P), & M + 2 \leq i < M + N \\ \frac{q_{i-1}}{P}(1 - A_{i-1} - P), & i = M + N \end{cases} \quad (4)$$

The relationships defined in (3) and (4) allow for the expression of the total probability in terms of the probability of being in the first inactive state, q_1 . This is useful because it

allows for the estimation of the probability of being in the state that reports simulated photon detections. By cascading the relations expressed in (4), and plugging into (3) gives (5),

$$q_1 = \left(1 + M + \left(\sum_{i=M+2}^{M+N-1} \prod_{j=M+1}^{i-1} (1 - A_j - P) \right) + \dots \right)^{-1} \cdot \left(\frac{1}{P} \prod_{k=M+1}^{M+N-1} (1 - A_k - P) \right) \quad (5)$$

Based on the above expression, the average count rate for a measurement with a given set of P , AP_0 , and α can be expressed as $q_I * f_s$. This can be seen by expressing q_I in terms of the probability of entering the state, equal to $q_{M+N}P + \sum_{i=M+1}^{M+N-1} q_i(A_i + P)$, which is equal to the probability of detecting a photon during a particular state residence time, given as $1/f_s$. Then by multiplying by the sampling rate of the simulation, f_s , we can estimate the count rate of the simulation in terms of counts per second. While there is not a unique mapping of measured count rate to detector characteristics, by comparing the estimated count rate from the steady state analysis and the count rate of the measurements, the set of appropriate detector characteristics can be narrowed, reducing the search size in the variable space.

To explore the dynamic properties of the detector response, we consider a time series of the detection state, given as $q_I[n]$. In the previous section we defined the probability of being in a state S_i in the detector steady state condition as q_i . In the next section, to allow for dynamics of the detector, we describe the probability of being in a state S_i at a particular time index in terms of clock cycles, n , as $q_i[n]$. The form of the discrete time autocorrelation for the system is given as $g_2[k] = \frac{\langle q_1[n]q_1[n+k] \rangle}{\langle q_1 \rangle^2}$. To simplify the calculation

of the autocorrelation, we modify the model by adding a known detection at $n=0$ in the form of a Kronecker delta. By arbitrarily setting $n=0$ when $q_I=1$, the autocorrelation of the signal will be equal to the expected probability of being state S_1 at a time lag of k . To determine this, we can describe the probability of being in state S_i at a certain time by a set of difference equations, given below in (6) and (7).

$$q_1[n] = \delta[n] + \left(\sum_{i=M+1}^{M+N-1} (A_i + P)q_i[n-1] \right) + \dots + Pq_{M+N}[n-1] \quad (6)$$

$$q_i[n] = \begin{cases} q_1[n-i+1], \dots & 1 \leq i < M+2 \\ q_1[n-i+1] \prod_{j=M+1}^{i-1} (1-A_j-P), \dots & M+2 \leq i < M+N \\ q_1[n-i+1] \prod_{j=M+1}^{i-1} (1-A_j-P) + \dots & \\ (1-P)q_i[n-1], i = M+N & \end{cases} \quad (7)$$

Equation 6 describes the transition from detecting a photon in the active period of the detector to the detection state, and (7) describes the transition of probability through the states in terms of the probability of being in the detection state, S_1 , at some previous point in time. Like in the steady state analysis, we can see a recurrence relationship that occurs in the final active state. To solve this system of equations for $q_i[n]$, we first take the z-transform of the above equations, given below in (8) and (9).

$$Q_1(z) = 1 + z^{-1} \left(\sum_{i=M+1}^{M+N-1} (A_i + P)Q_i(z) \right) + PQ_{M+N}(z) \quad (8)$$

$$q_i[n] = \begin{cases} z^{-(i-1)}Q_1(z), \dots & 1 \leq i < M+2 \\ z^{-(i-1)}Q_1(z) \prod_{j=M+1}^{i-1} (1-A_j-P), \dots & M+2 \leq i < M+N \\ \frac{z^{-(M+N-1)}Q_1(z)}{1-(1-P)z^{-1}} \prod_{j=M+1}^{M+N-1} (1-A_j-P), \dots & \\ i = M+N & \end{cases} \quad (9)$$

Plugging in the expressions from (9) into (8) yields the following solution for $Q_1(z)$,

$$Q_1(z) = \left(\sum_{i=0}^1 a_i z^{-i} \right) \left(\sum_{j=0}^{M+N} b_j z^{-j} \right)^{-1}, \quad (10)$$

where the coefficients of a_i and b_j can be expressed as;

$$a_i = \begin{cases} 1, & i = 0 \\ -(1-P), & i = 1 \end{cases} \quad (11)$$

$$b_j = \begin{cases} 1, & j = 0 \\ -(1 - P), & j = 1 \\ 0, & 1 < j < M + 1 \\ -(A_j + P), & j = M + 1 \\ (1 - P)(A_{j-1} + P) - \dots \\ \quad (A_j + P)(1 - A_{j-1} - P), & j = M + 2 \\ (1 - P)(A_{j-1} + P) \prod_{k=M+1}^{j-2} (1 - A_k - P) - \dots \\ (A_j + P) \prod_{k=M+1}^{j-1} (1 - A_k - P), & M + 2 < j < M + N \\ (1 - P)(A_{j-1} + P) \prod_{k=M+1}^{j-2} (1 - A_k - P) - \dots \\ P \prod_{k=M+1}^{j-1} (1 - A_k - P), & j = M + N \end{cases} \quad (12)$$

Finally, by taking the inverse z-transform of $Q1(z)$, we are left with the estimated autocorrelation of the detector for a given set of properties (P , AP_0 , and α) to compare to the measured autocorrelation. We fit for the properties by minimizing a loss function of the form, $L(P, AP_0, \alpha) = \sum (\widehat{g}_2[k] - g_2[k])^2 + \lambda R(P, AP_0, \alpha)$, where $g_2[k]$ is the measured autocorrelation function, $\widehat{g}_2[k]$ is the estimated autocorrelation function calculated as

$$\widehat{g}_2[k] = \frac{q_1[k]}{\langle q_1[k] \rangle}, R \text{ is a regularization term that is based on the estimated count rate of}$$

the measurement, and λ is the scaling factor for the regularization term. Below an empirically chosen relative difference, d_r , in count rate, $R = 0$. Above the relative difference threshold, detector characteristics are penalized and the regularization term takes the form, $R = (|\widehat{CR} - CR| - d_r CR)^2$, where CR is the measured count rate, \widehat{CR} is the estimated count rate calculated as $\widehat{CR} = q_1 * f_s$. For this work the relative difference threshold was set to 0.1.

With this model in place, we can characterize the performance of the detector during the different operating conditions in the terms discussed above. To then determine the effects of the detector characteristics on the cross-correlation function and the fitted BF_i , we develop a model which generates synthetic data to allow for estimates of corrections to the measured BF_i . To generate these synthetic data, we first define the three types of photon detection events (real photon detections arising from the back scattered light, afterpulsing counts, and thermal dark counts). To simulate the real intensity signal, $I[n]$, white noise is convolved with the electric field autocorrelation function, $g_1(\tau)$, corresponding to the estimated optical properties of the sample (μ_a , μ_s' , index of refraction), the wavelength of light (λ), the source-detector separation (ρ), and the effective D_b value. The signal is then normalized such that $\langle I[n] \rangle = \sqrt{\langle (I[n] - \langle I[n] \rangle)^2 \rangle}$. This corresponds to an intensity signal where the coherence factor, β , is equal to 1, indicating we are measuring the fluctuation of a single speckle. In

practice for the reflectance geometry, β is limited to 0.5, as two independent polarizations are captured by the single mode fiber.

To account for this, we generate two instances of $I[n]$ and add the two together. We then convert the generated intensity signal to a probability of photon detection within a certain time bin, n , given as $P_S[n] = \frac{I[n]}{\langle I[n] \rangle} * \frac{CR}{f_s}$, where $P_S[n]$ is the probability of detecting a signal photon in bin n , CR is the count rate of the measurement in counts per second, and f_s is the sampling rate of the simulation. To simulate the afterpulsing of the detector, the afterpulsing parameters estimated from the real measurement are used describe the probability of detection, equal to $P_{AP}[n - n_0] = \frac{AP_0}{f_s} \left(\frac{n - n_0}{M + 1} \right)^{-\alpha}$, where n_0 is the time bin when the last photon was detected. The dark counts are estimated as a uniform probability across all time bins n , equal to $P_{DCR} = \frac{DCR}{f_s}$, where DCR is the dark count rate in counts per second. The total probability for photon detection during a particular time bin is given as $P[n] = P_S[n] + P_{AP}[n - n_0] + P_{DCR}$. Random numbers, $R[n]$, are sampled from a uniform distribution spanning from 0 to 1 for each time bin in order until $P[n] > R[n]$, indicating a photon detection event. The timestamp of the detected event is saved and the detector is then set to the “inactive state” by skipping the next M time bins, corresponding to the hold off time. This process is repeated until a sufficient number of photon detection events are recorded to generate the intensity autocorrelation function, $g_2(\tau)$. To simulate the use of cross-correlation, the model is run again using the same intensity signal, as though another detector is being used. The seed of the RNG was reset at the beginning of each set of simulations to ensure the observations of the speckle signal for each detector were unique. This model is shown pictorially in Fig. 2. Simulated $g_2(\tau)$'s calculated by cross-correlation are then simulated for a range of D_b values to determine the calibration curve to convert the apparent D_b to true D_b . This calibration is dependent only on the detector characteristics and applicable across all samples, with the caveat that the distortion of the correlation curve is dependent upon the measurement count rate, as is explored in further sections. To ensure accurate recovery of BF_i from the experimental data, calibration curves are generated for each calculated $g_2(\tau)$ to account for possible differences in measurement count rate.

III. Results and discussion

A. Comparison of autocorrelation and cross-correlation

To show the effects on $g_2(\tau)$ caused by calculating either the autocorrelation of the summed detector streams or the cross-correlation of the detector streams, we plot the correlation functions against the reference measurement correlation function taken by the SNSPD from the CW intralipid phantom measurements in Fig. 3(a) and Fig. 3(b). The obvious artifacts present in the autocorrelation, including the peaking behavior and oscillations that can be attributed to the afterpulsing of the detector as well as the step function drop at the hold off time, would likely prevent accurate characterization of the BF_i . The cross-correlation curve is relatively clean, with the major features present in the autocorrelation now absent, as seen in Fig. 3(a). When compared to the reference measurement in Fig. 3(b), it can be seen that the decorrelation of the InGaAs/InP detectors is slowed, the possibility of which

was discussed in the sections above due to the presence of afterpulsing counts remaining in the detector streams. This effect is not limited to the cross-correlation calculation, as the can be seen in Fig. 3(b). Although the major slowing factor of the autocorrelation is the afterpulsing tail, even at long lags where afterpulsing should be resolved ($\tau > 1$ ms), the autocorrelation still decays slower than the reference measurement. To show the effects of the slowing, BF_1 values are fit from cross-correlation functions calculated from each of the detector setting manipulations and plotted as a box plot in Fig. 3(c). The BF_1 fit from the InGaAs/InP detectors can be seen to be $\sim 30\%$ lower than the reference measurement. The large error in the accuracy of the absolute value provides the motivation for the use of the models described in the methods section. Even though a large discrepancy in BF_1 is present at this stage, the use of cross-correlation removes the bulk of the artifacts present when the autocorrelation is calculated, and provides promising results.

B. Estimation of detector characteristics at different operating conditions

The first step to correcting the BF_1 estimated from the cross-correlation $g_2(\tau)$ is the determination of the detector characteristics (P , AP_0 , and α) corresponding to each operating condition (excess bias voltage, hold off time, and temperature). An example of the fitted autocorrelation curves from each individual SPAD can be seen below in Fig. 4(a). Here we plot the average of 5 modeled autocorrelation functions generated by multi-starting the optimization with plus and minus a standard deviation in the shading. We can see from this figure that the model provides an autocorrelation function that matches the shape of the afterpulsing corrupted correlation function. Further, we can also appreciate the differences in the correlation curve for the detectors that are being operated in the same conditions, indicating a fair amount of variable performance ($P = [8.53, 5.71]$ kcps, $AP_0 = [175.9, 160.7]$ kcps, $\alpha = [1.79, 2.10]$, for detectors 1 and 2, respectively). To demonstrate that the fitted parameters of the model are consistent with the expected results for these SPAD detectors, e.g. increasing the excess bias voltage will increase the dark count and afterpulsing, we plot the estimated detector characteristics (P , AP_0 , and α) for one of the detectors operated in free running mode at a single temperature in Fig. 4(c), 4(d), and 4(e) and at a single excess bias voltage in Fig. 4(f), 4(g), and 4(h). The estimated detector characteristics match intuition, with afterpulsing probability increasing with increasing bias voltage and decreasing hold off time. The decay rate of the afterpulsing doesn't show consistent behavior with bias voltage, though as was expected, the increasing magnitude of the parameter with increasing hold off time indicates a faster decay of the afterpulsing. For the background count rate, we see consistent increases in the probability of detection with increasing bias voltage, but the relationship with the hold off time is less clear. Though we would expect the probability of detection in the active state to be equal across the different hold off times, for shorter hold offs, the estimated parameter is lower than for the longer hold off times, where for hold off times > 30 μ s, the curves overlap, seen in Fig. 4(d) and 4(g). We believe these results may be due to the relatively extreme afterpulsing that is present in the short hold off time operating conditions. We also see results that are consistent with expectation as the temperature of the SPAD is varied. We see decreases in the afterpulsing rate with increased temperature as well as having an increase in the count rate. The decay rate of the afterpulsing again shows a relatively inconsistent behavior with the change in detector operating parameter, though with increasing hold off time, the

rate of the afterpulsing decay increases. These results provide evidence that the detector characteristics can be determined across a range of operating conditions. Finally, by using the fitted detector characteristics, we can correct the BF_i derived from the cross-correlation and show that the BF_i estimated from all free running conditions explored match well with the BF_i derived from the reference measurements, shown in Fig. 4(b).

C. Signal-to-noise ratio at different operating conditions

For the selection of optimal operating parameters, we use the SNR of the plateau of the measured cross-correlation $g_2(\tau)$. Below we compare surface plots for the CW operation (Fig. 5(a)) and gated operation (Fig. 5(b)) for the SNR of the plateau of $g_2(\tau)$ as a function of the hold off time and excess bias voltage and temperature. For both sets of results, we calculate $g_2(\tau)$ at a frequency of 0.1 Hz from the intralipid phantom data. To simplify the plotting of the results, we collapse the data along the temperature axis and plot the SNR results for the CW and gated operation in Fig. 5(c) and 5(d), respectively. From these results, it is readily evident that gated operation, even when measurements are not time resolved, allow for much greater SNR, likely due to the reduction in the duty cycle of the dark count rate and the possibility of having afterpulses. Another beneficial feature of gated operation is the increase in SNR with increasing bias voltage, indicating that the increases in detection efficiency are not being overwhelmed by increases in the noise counts, as we theorize is the case for CW operation. Based on the results of these experiments, only gated operation was used for the *in vivo* experiments, with the SNR of the CW measurement not being high enough. While the SNR of the curve seems sufficient to fit for BF_i , the combination of the higher sampling rate and the increase in the variability of fitted BF_i after correction imposes the need for a slightly higher SNR than is traditionally required to make measurements. The settings used for the *in vivo* experiments were determined to be a temperature of 225K, an excess bias of 7V and a hold off time of 5 μ s.

D. In vivo tourniquet tightening experiment and correction of the measured BF_i

With the optimal operating conditions determined for these detectors, measurements of *in vivo* blood flow were taken. The time trace of BF_i fit from the reference detector, naively fit BF_i from the InGaAs/InP detector cross-correlation, and model corrected BF_i from the InGaAs/InP detector cross-correlation can be seen in Fig. 6(a). In these experiments, the hyperemia peak observed at the release of the tourniquet was relatively blunted, though this could be due to the appropriateness of the model used to fit the data at a relatively short source-detector separation [35] or the relatively small reduction in blood flow (~30%) caused by the tourniquet. Relative changes in BF_i normalized to the baseline period show similarity between detectors (not plotted here), though as is seen in the next section, the relationship between the true BF_i and the apparent BF_i is nonlinear, which could complicate the use of relative BF_i . To correct for the offset in absolute BF_i , we apply the synthetic data model to determine the relationship between the estimated BF_i and true BF_i for the set of detector characteristics determined by fitting the individual detector autocorrelation function. Seen below in Fig. 6(b), we show a scatterplot of BF_i values fit from the reference measurement $g_2(\tau)$ and the InGaAs/InP cross-correlation $g_2(\tau)$. For the uncorrected data, the BF_i is significantly lower, with the discrepancy becoming more severe for higher values

of true BF_i . Following correction, the data falls along the line of unity, and accurate BF_i estimates can be made using the afterpulsing corrupted $g_2(\tau)$.

E. Estimating BF_i distortion for a more realistic source-detector separation optimizing for cerebral sensitivity

In this work, all InGaAs/InP SPAD measurements were made at relatively short source-detector separations (~ 1 cm) due to the photon detection efficiency of these particular modules, which are currently optimized to be most sensitive to light around 1500 nm [36]. While we were limited to the short separation range in this work due to detector fiber coupling inefficiencies, we can still estimate the performance of a SPAD with comparable afterpulsing as though measurements were made at longer separations. Based on previous *in vivo* measurements of photon counts at a long source-detector separation taken with an SNSPD detector and normalizing that count rate by the detector efficiency of the free space version of this InGaAs/InP SPAD ($\sim 40\%$ @ 1064nm), we estimate we should collect between 5 kcps and 50 kcps. Plugging these estimated count rates into the synthetic data model yields the results seen below in Fig. 7(a) and 7(b), comparing the true BF_i and the BF_i estimated from the naively fit autocorrelation function. The BF_i underestimation strongly depends on count rate, and for the low count rates explored below, the degree of slowing matches the slowing seen in the in-vivo results where the count rate was 8.5 kcps total between the two detectors. Comparing to the results presented above in Fig. 6(b), the distortion to the recovered BF_i is greater with the faster decorrelation seen at the longer source-detector separation, though with count rates >10 kcps the relationship between true BF_i and recovered BF_i does not “saturate” in a physiologic range of blood flow index, and by applying our distortion correction model, will make the recovery of the blood flow index possible.

IV. Conclusion

Here we have detailed a set of methods to correct the inherent shortcomings of InGaAs/InP sPAD detectors when applied to diffuse correlation spectroscopy. Through a combination of detector cross-correlation and appropriate modeling, distortions of the correlation function can be accounted for and accurate BF_i values can be measured. To start we show that using cross-correlation, InGaAs/InP detectors become usable for DCs despite the strong AP and long holdoff time. A residual afterpulsing artifact is still present at low count rate, and we provided a detector modeling approach to remove the systematic bias from BF_i values derived from the cross-correlation data. We showed that accurate characterization of detector effects was possible across a range of operating parameters. Although here we assume a power law model for the afterpulsing, inclusion of different types of models for afterpulsing, as have been explored in the literature [37], is easily possible by swapping the function dependence of A_i . The use of these models also seems to give results similar to those made through time-correlated single photon counting experiments without the need for the required instrumentation, which could be a benefit in detector characterization [38]. In general, for detectors with this profile of afterpulsing and detection efficiency, we find that operating the detector in gated mode with a high excess bias voltage provides the best performance, allowing for *in vivo* measurements to be made. We believe the application

of this analysis can open the door for higher sNR DCs measurements made at longer wavelengths.

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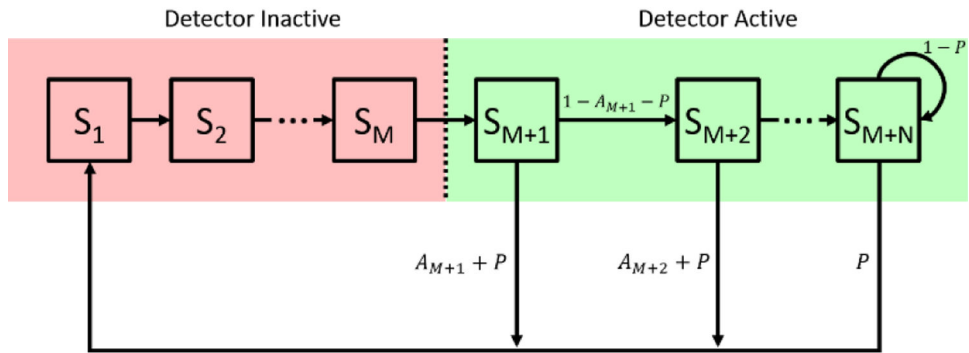


Fig. 1. State space model used to represent a detector module. Detector states are labeled with respect to their position relative to the detection state S_1 , with the same numbering given for their state transition probabilities, which are labeled near the arrow indicating the state transition.

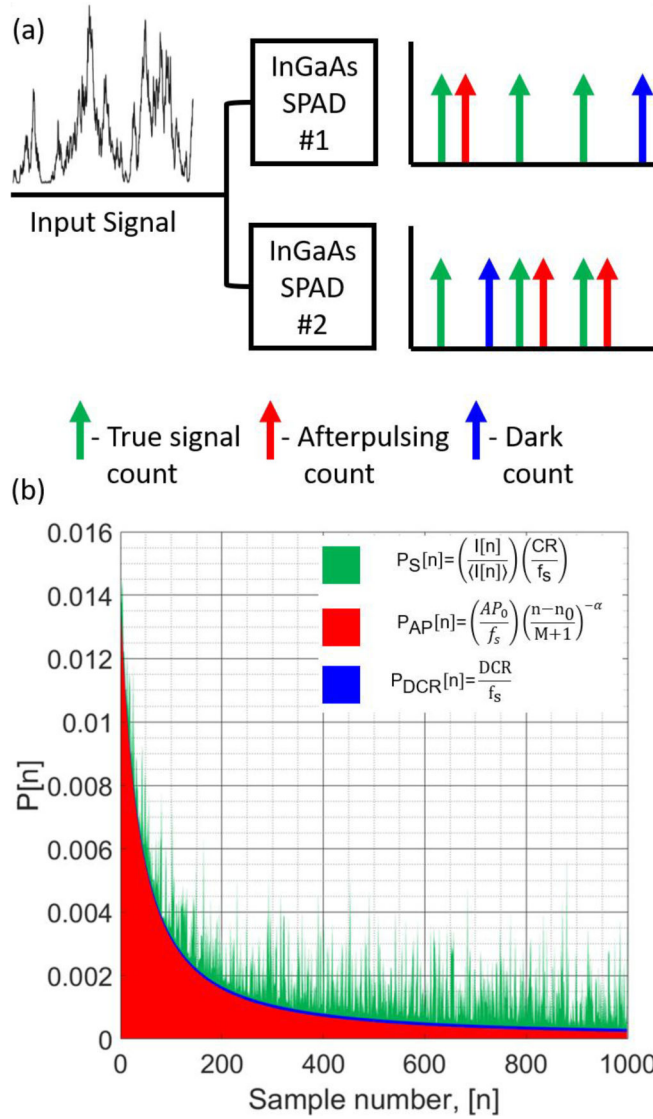


Fig. 2. (a) Pictorial model description for the synthetic data model. The photon counts collected by the simulated detectors are shown with different types of counts color coded to their descriptions. (b) An example of the probabilities for the detection of each type of count is shown color coded in the same way as in (a). Inputs to the model, including the count rate of the back scattered light, DCR of the detector, and afterpulsing probabilities ($P=10$ keps, $AP_0=140$ keps, and $\alpha=1.37$), are used to determine the probabilities.

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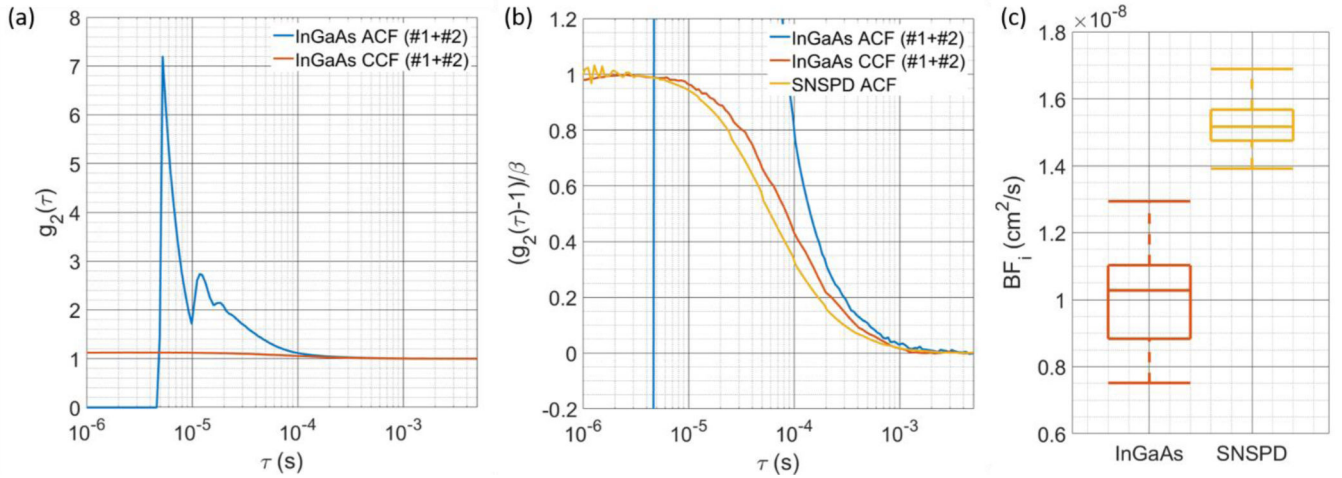


Fig. 3.

(a) Comparison of the $g_2(\tau)$ calculated with autocorrelation of the combined counts of both detectors and the cross-correlation between the counts of each detector. The large artifacts due to afterpulsing are present in the autocorrelation calculation, though cross-correlation can be seen to remove the artifacts and extend the g_2 to times below the hold-off time.

(b) The comparison between the SNSPD autocorrelation, InGaAs/InP autocorrelation, and InGaAs/InP cross correlation shows that the cross-correlation is not completely without distortion, showing a slower decay. Quantification of this decay is shown in (c), where BF_1 is fit from $g_2(\tau)$'s calculated from measurements at different operating conditions. The consistently lower BF_1 is likely due to the “extension” of correlation by afterpulsing counts.

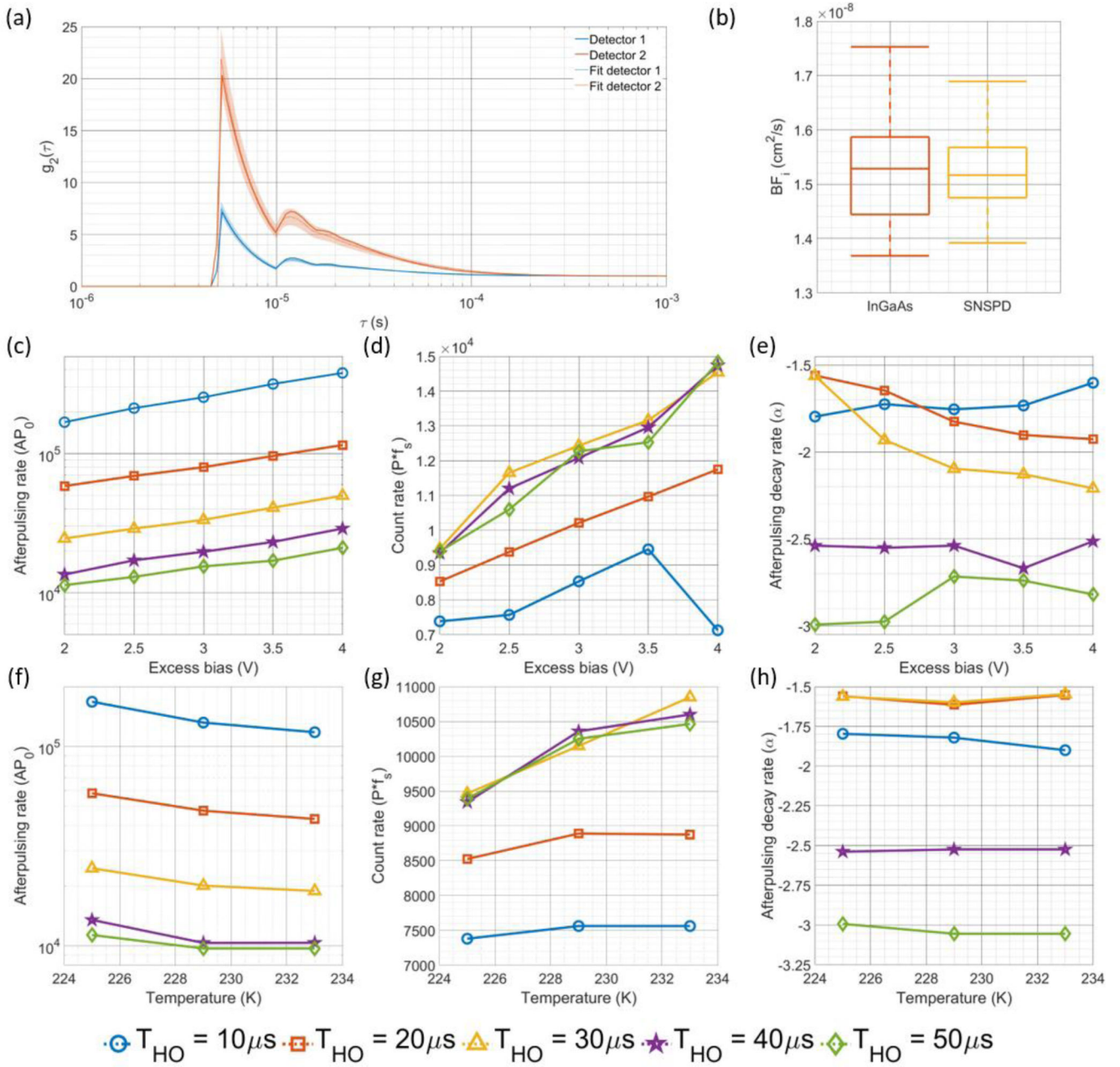


Fig. 4.

(a) Comparison of the autocorrelation calculated from the individual detector counts and their respective estimated autocorrelations fit using the detector model for a single set of detector operating parameters (Excess bias = 2.5 V, Hold off time = 10 μ s, Temperature = 229 K). (b) Using the fitted properties, the BFI derived from the cross-correlation is corrected to account for the slowing of the correlation decay caused by the afterpulsing, and can be seen to match the BFI derived from the reference measurement. Fitting the correlation curves with the state space model allows for estimation of the afterpulsing probability at hold off time, shown in (c) and (f), measured count rate, shown in (d) and (g), and afterpulsing decay rate, shown in (e) and (h). The plots in (c), (d), and (e) are shown for a

fixed temperature of 225K, and for the plots in (f), (g), and (h), the excess bias voltage is fixed at 2.0 V. The fit values for these parameters generally follow the intuition surrounding how altering the operating parameters (excess bias voltage, hold off time, and temperature) should affect the detector characteristics (P , AP_0 , and α). Based on the count rate estimation shown in (d), at extremely high afterpulsing rates/detector non-linearity conditions like those for the shorter hold off time condition, the estimation of count rate might be inaccurate, as shown in the difference between the 10 and 20 μ s hold off time vs. the rest of the measurements.

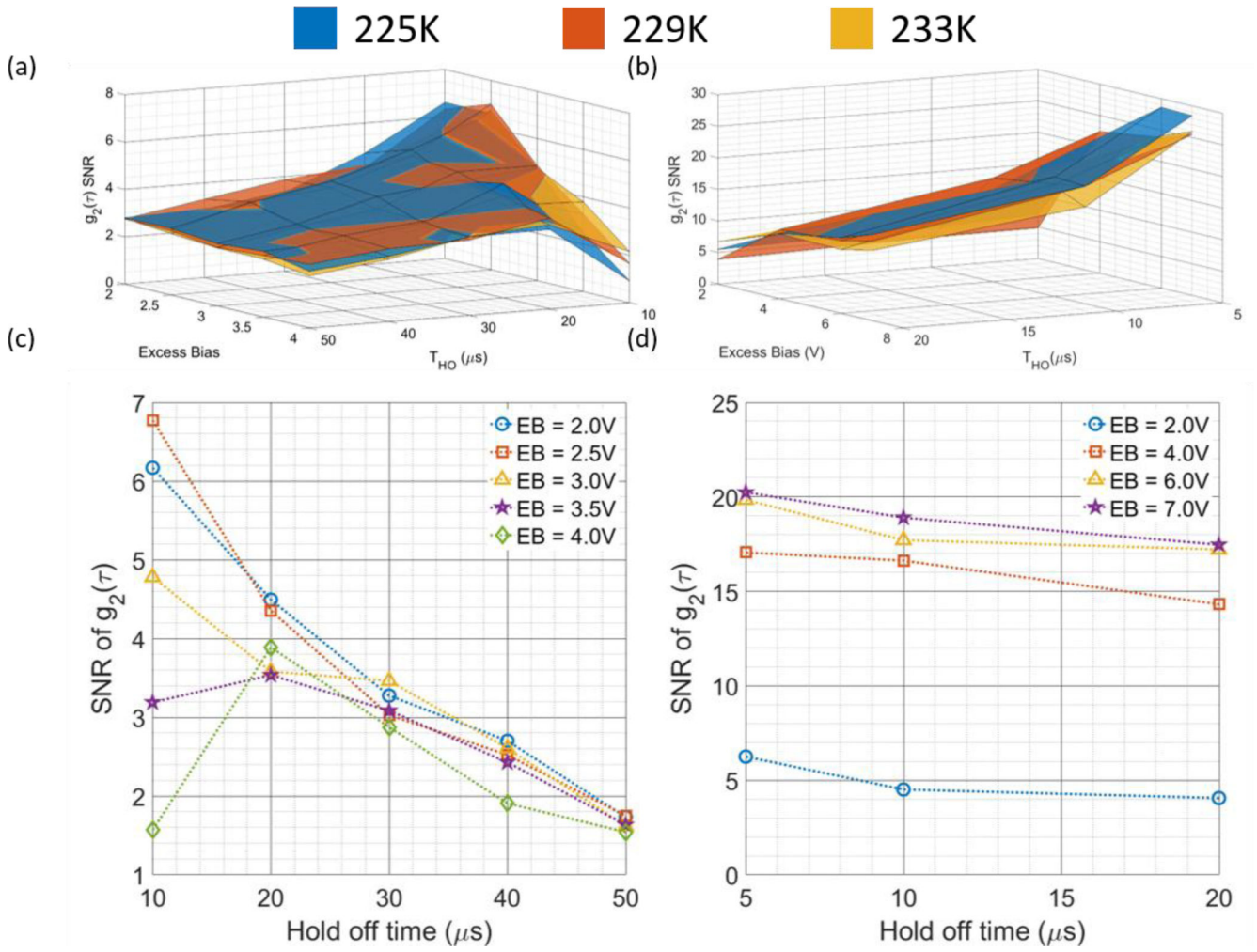


Fig. 5. The comparisons of the SNR of the $g_2(\tau)$ curves as a function of the three operating parameters (excess bias voltage, hold off time, and temperature) are shown in (a) for the CW operation and (b) for the gated operation. To increase readability, the results present in (a) and (b) are averaged by the temperature, and presented in (c) and (d), respectively. It can be seen that for CW operation, a high bias voltage does not improve SNR, indicating that the noise counts grow faster than the increase in signal counts mediated by an increase in detection efficiency. For gated operation, the lower duty cycle allows for a reduction in dark counts, and the increase in detection efficiency is better realized, as indicated by the increased SNR at higher bias voltage. In general, across these measurements, increases in the hold off time are detrimental to SNR as they reduce the duty cycle of the detector and limit the number of back scattered photons that can be detected.

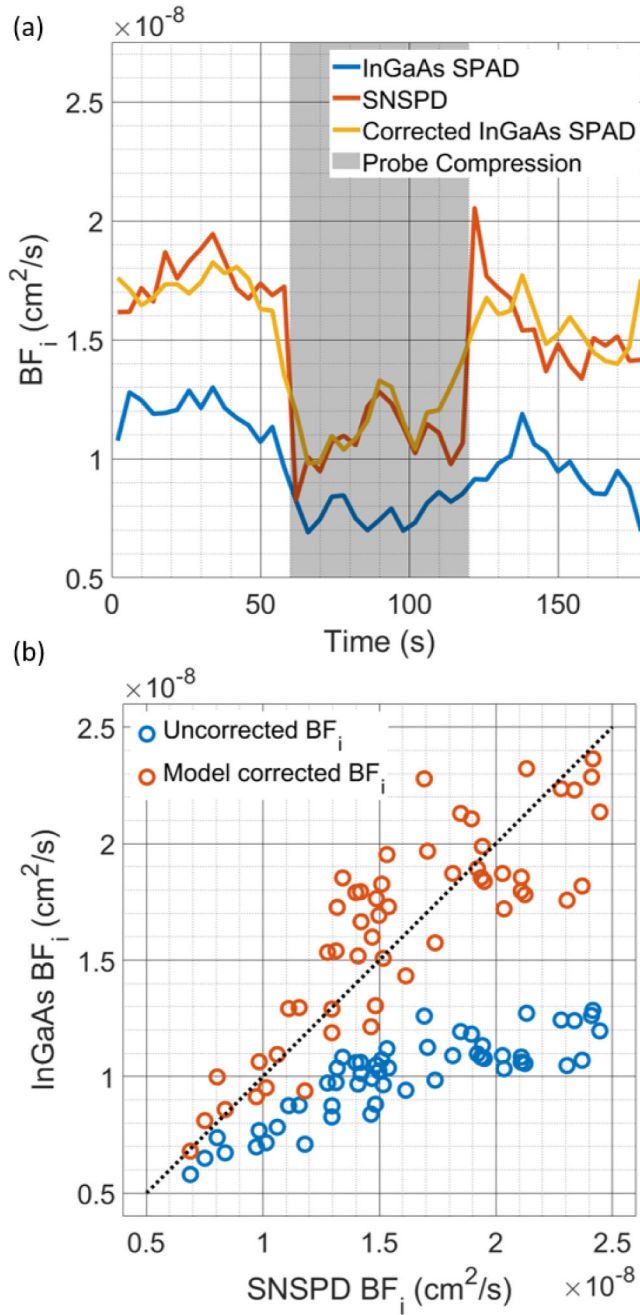


Fig. 6. (a) Comparison of the time course of fitted BF_i during the probe compression protocol. The effect of the afterpulsing on the BF_i is seen in the lower BF_i fitted for the uncorrected data, but following correction, the InGaAs/InP detector cross-correlation gives very similar results to that of the SNSPD. (b) Comparison of the BF_i fitted from the SNSPD and InGaAs/InP measurements with and without the correction by the model. Large systematic errors can be seen in the estimate of BF_i without correction, but the model allows the extraction of accurate BF_i values. Due to the compression of the dynamic range of the autocorrelation

decay at higher BF_1 values, the variability of the BF_1 estimates after correction can be seen to increase with the degree of the discrepancy between the true BF_1 and the naively fit BF_1 .

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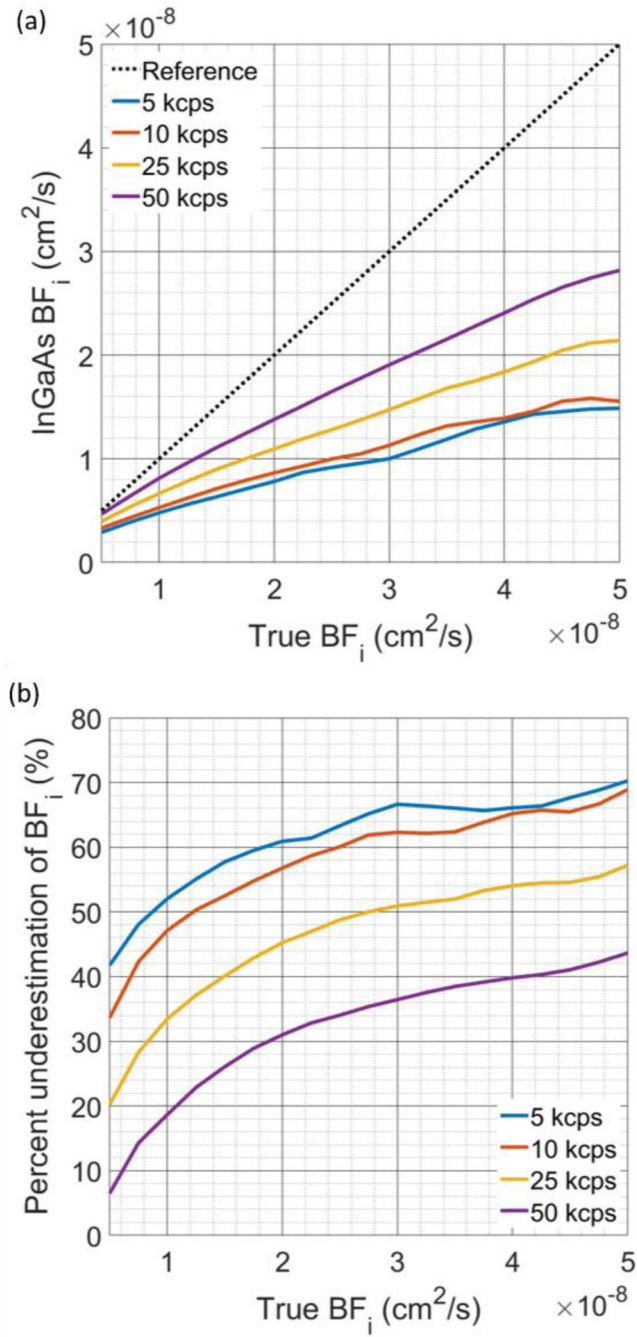


Fig. 7:

(a) Comparison of the recovered BF_i from $g_2(\tau)$ curves generated by the synthetic data model. The detector characteristics for each detector used for these simulations were $AP_0 = [180, 140]$ kcps, $DCR = [0.25, 0.25]$ kcps, and $\alpha = [1.37, 1.54]$, as fit from the measurements made at the operating conditions used for the *in vivo* experiments. At low count rates, the BF_i recovered from the curves shows more severe underestimation, seen in both (a) and (b), though with a higher photon count rate, the degree of underestimation is improved. The relationship shown provides a rationale for the relative BF_i matching

well between the InGaAs/InP measurement and the reference measurement, as these measurements were made over a relatively small range of BF_i values. With an increase in the range of BF_i made during a particular measurement, this accuracy would likely be degraded, as the relationship between InGaAs/InP derived BF_i and true BF_i is nonlinear.

TABLE I

Detector operating parameters

	Free running	Gated
Excess bias voltage	2.0 V, 2.5 V, 3.0 V, 3.5 V, 4.0 V	2.0 V, 4.0 V, 6.0 V, 7.0 V
Hold off time	10 μ s, 20 μ s, 30 μ s, 40 μ s, 50 μ s	5 μ s, 10 μ s, 20 μ s
Temperature	225 K, 229 K, 233 K	

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