Appendix A: Proof of Theorem 1

Based on the analysis in Section 2.2, the proposed NTOA scheme (6) can be reformulated as

$$J\dot{\theta} = \dot{r}_{\mathrm{d}} - k_{\mathrm{P}}(f(\theta) - r_{\mathrm{d}}) - k_{\mathrm{I}} \int_{0}^{t} (f(\theta) - r_{\mathrm{d}}) \mathrm{d}\tau.$$

which is equivalent to

$$\dot{e}(t) = -k_{\rm P}e(t) - k_{\rm I} \int_0^t e(\tau)d\tau,$$

where $e(t) = f(\theta) - r_{\rm d}$ and $\dot{e}(t) = J\dot{\theta} - \dot{r}_{\rm d}$ with $J = \partial f(\theta)/\partial \theta$. (8) is a compact vector form of the following set of m decoupled equations:

$$\dot{e}_i(t) = -k_{\rm P}e_i(t) - k_{\rm I} \int_0^t e_i(\tau) d\tau,$$

in which $i = 1, \dots, m$. To analyze the *i*th subsystem (9), the following Lyapunov function candidate [38, 39] is defined:

$$v_i(t) = e_i^2(t) + k_{\rm I} \left(\int_0^t e_i(\tau) d\tau \right)^2.$$

Evidently, $v_i(t) > 0$ for any $e_i(t) \neq 0$ or $\int_0^t e_i(\tau) d\tau \neq 0$, and $v_{ij}(t) = 0$ only for $e_{ij}(t) = \int_0^t e_{ij}(\tau) d\tau = 0$. This guarantees the positive-definiteness of the Lyapunov function candidate $v_i(t)$. For this Lyapunov function candidate, the time derivative $\dot{v}_i(t)$ can be derived as

$$\dot{v}_i(t) = \frac{dv_i(t)}{dt} = 2e_i(t)\dot{e}_i(t) + 2k_{\rm I}e_i(t)\int_0^t e_i(\tau)d\tau = -2k_{\rm P}e_i^2(t) \le 0,$$

which guarantees the negative-definiteness of $\dot{v}_i(t)$. On the basis of the Lyapunov theory, the trajectory of the Cartesian error e(t) for the proposed NTOA scheme (6) can be concluded to be asymptotically stable. The proof is thus completed.

Appendix B: Proof of Theorem 2

Let us define $\varepsilon(t) = \int_0^t e(\tau) d\tau \in \mathbb{R}^m$. Then, $\dot{\varepsilon}(t) = e(t)$ and $\ddot{\varepsilon}(t) = \dot{e}(t)$. The *i*th subsystem (9) is thus rewritten as follows:

$$\ddot{\varepsilon}_i(t) = -k_{\rm P}\dot{\varepsilon}_i(t) - k_{\rm I}\varepsilon_i(t),$$

where $\varepsilon_i(t)$, $\dot{\varepsilon}_i(t)$, and $\ddot{\varepsilon}_i(t)$ are the *i*th elements of $\varepsilon(t)$, $\dot{\varepsilon}(t)$, and $\ddot{\varepsilon}(t)$, respectively.

As defined previously, the design parameters $k_{\rm P}$ and $k_{\rm I}$ satisfy $k_{\rm P}^2 > 4k_{\rm I}$ numerically. Therefore, let $\vartheta_1 = (-k_{\rm P} + \sqrt{k_{\rm P}^2 - 4k_{\rm I}})/2$ and $\vartheta_2 = (-k_{\rm P} - \sqrt{k_{\rm P}^2 - 4k_{\rm I}})/2$. Then, considering that $\varepsilon_i(0) = 0$ and $\dot{\varepsilon}_i(0) = e_i(0)$, the analytical solution to (10) is formulated as

$$\varepsilon_i(t) = \frac{e_i(0)(\exp(\vartheta_1 t) - \exp(\vartheta_2 t))}{\sqrt{k_P^2 - 4k_I}}.$$

Because $e(t) = \dot{\varepsilon}(t)$, the following result is further obtained:

$$e_i(t) = \frac{e_i(0)(\vartheta_1 \exp(\vartheta_1 t) - \vartheta_2 \exp(\vartheta_2 t))}{\sqrt{k_P^2 - 4k_I}}.$$

The vector-form Cartesian error e(t) is thus derived as follows:

$$e(t) = \frac{e(0)(\vartheta_1 \exp(\vartheta_1 t) - \vartheta_2 \exp(\vartheta_2 t))}{\sqrt{k_P^2 - 4k_I}}.$$

On the basis of the above analysis (as well as the proof of Theorem 1), starting from an initial state $e(0) \neq 0$ (i.e., a nonzero initial error), the trajectory of the Cartesian error e(t) for the proposed NTOA scheme (6) can be concluded to converge to zero exponentially. In other words, the Cartesian error e(t) synthesized by (6) has the property of exponential convergence, which now completes the proof.

Appendix C: Proof of Theorem 3

Based on Appendix A, the noise-polluted NTOA scheme (7) can be reformulated as follows:

$$J\dot{\theta} = \dot{r}_{\mathrm{d}} - k_{\mathrm{P}}(f(\theta) - r_{\mathrm{d}}) - k_{\mathrm{I}} \int_{0}^{t} (f(\theta) - r_{\mathrm{d}}) \mathrm{d}\tau + \delta(t),$$

which is equivalent to

$$\dot{e}(t) = -k_{\mathrm{P}}e(t) - k_{\mathrm{I}} \int_0^t e(\tau)d\tau + \delta(t).$$

By using the Laplace transformation [41], the ith subsystem of (11) is as follows:

$$se_i(s) - e_i(0) = -k_P e_i(s) - \frac{k_I}{s} e_i(s) + \delta_i(s).$$

Then, the following result is obtained:

$$e_i(s) = \frac{s(e_i(0) + \delta_i(s))}{s^2 + k_P s + k_i}.$$

Evidently, the transfer function is derived as follows:

$$\frac{s}{s^2 + k_{\rm P}s + k_{\rm I}},$$

of which the poles are $(-k_P + \sqrt{k_P^2 - 4k_I})/2$ and $(-k_P - \sqrt{k_P^2 - 4k_I})/2$. Note that, as defined previously, k_P and k_I satisfy $k_P^2 > 4k_I > 0$ numerically.

Therefore, such two poles are located on the left half-plane, indicating the stability of the system (12).

Note that $\delta(t) = c \in \mathbb{R}^m$ (being a constant), and thus $\delta_i(s) = c/s$. For (12), using the final value theorem [41] yields

$$\lim_{t \to \infty} e_i(t) = \lim_{s \to 0} s e_i(s) = \lim_{s \to 0} \frac{s^2(e_i(0) + c/s)}{s^2 + k_P s + k_I} = 0.$$

According to (13), $\lim_{t\to\infty} \|e(t)\|_2 = 0$ with symbol $\|\cdot\|_2$ denoting the two norm of a vector. On the basis of the above analysis, the trajectory of e(t) for the noise-polluted NTOA scheme (7) can be concluded to converge to zero. In other words, the Cartesian error e(t) synthesized by (7) has a convergence property, which now completes the proof.

Appendix D: Proof of Theorem 4

As presented in Appendix C, the noise-polluted NTOA scheme (7) can be rewritten as follows:

$$\dot{e}(t) = -k_{\mathrm{P}}e(t) - k_{\mathrm{I}} \int_0^t e(\tau)d\tau + \delta(t),$$

where the ith $(\forall i = 1, \dots, m)$ subsystem is formulated as

$$\dot{e}_i(t) = -k_{\rm P}e_i(t) - k_{\rm I} \int_0^t e_i(\tau)d\tau + \delta_i(t).$$

Considering that $k_{\rm P}$ and $k_{\rm I}$ satisfy $k_{\rm P}^2 > 4k_{\rm I} > 0$ numerically, the solution to (14) is obtained as follows:

$$e_i(t) = \frac{e_i(0)(\vartheta_1 \exp(\vartheta_1 t) - \vartheta_2 \exp(\vartheta_2 t))}{\sqrt{k_P^2 - 4k_I}} + \frac{1}{\sqrt{k_P^2 - 4k_I}} \int_0^t (\vartheta_1 \exp(\vartheta_1 (t - \tau))) - \vartheta_2 \exp(\vartheta_2 (t - \tau))) \delta_i(\tau) d\tau.$$

From the triangle inequality,

$$|e_i(t)| \leqslant \frac{|e_i(0)(\vartheta_1 \exp(\vartheta_1 t) - \vartheta_2 \exp(\vartheta_2 t))|}{\sqrt{k_{\rm P}^2 - 4k_{\rm I}}} + \frac{\int_0^t |\vartheta_1 \exp(\vartheta_1 (t - \tau))| |\delta_i(\tau)| \mathrm{d}\tau}{\sqrt{k_{\rm P}^2 - 4k_{\rm I}}} + \frac{\int_0^t |\vartheta_2 \exp(\vartheta_2 (t - \tau))| |\delta_i(\tau)| \mathrm{d}\tau}{\sqrt{k_{\rm P}^2 - 4k_{\rm I}}}.$$

It follows the above inequality that

$$|e_i(t)| \leqslant \frac{|e_i(0)(\vartheta_1 \exp(\vartheta_1 t) - \vartheta_2 \exp(\vartheta_2 t))|}{\sqrt{k_P^2 - 4k_I}} + \frac{2}{\sqrt{k_P^2 - 4k_I}} \max_{0 \leqslant \tau \leqslant t} |\delta_i(\tau)|.$$

Therefore, the following result is obtained:

$$\lim_{t \to \infty} \sup \|e(t)\|_2 \leqslant 2\sqrt{\frac{m}{k_{\mathrm{P}}^2 - 4k_{\mathrm{I}}}} \max_{0 \leqslant \tau \leqslant t} |\delta_i(\tau)|.$$

On the basis of the above analysis, the Cartesian error e(t) synthesized by the noise-polluted NTOA scheme (7) in the presence of bounded time-varying noise can be concluded to be bounded, with the steady-state error being bounded by $2\sqrt{m} \max_{0 \le \tau \le t} |\delta_i(\tau)|/\sqrt{k_{\rm P}^2 - 4k_{\rm I}}$. The proof is thus completed.